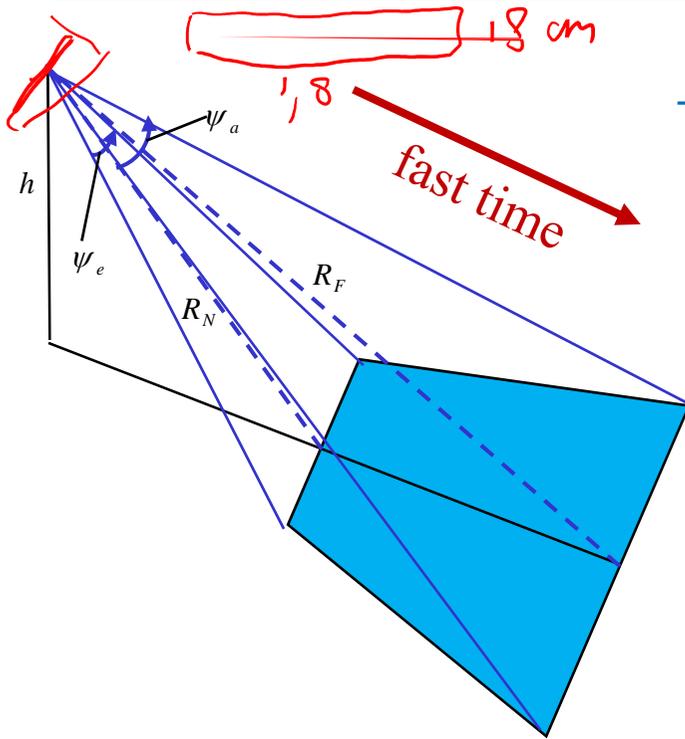

Synthetic Aperture Radar

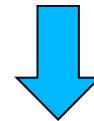
Pierfrancesco Lombardo

Single pulse radar echo



- Any desired value is achievable using a pulse with B large enough!

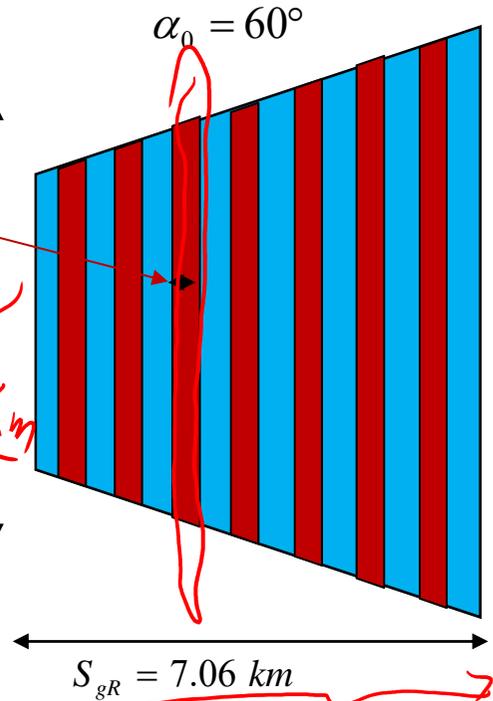
$$\delta_{gR} = \frac{k c}{\sqrt{3} B}$$



Example: B = 450 MHz
 k = 1.3 Hamming (PSL 43 dB)

$$\delta_{gR} = 0.5 \text{ m}$$

$$D_{s0} = 344 \text{ m}$$



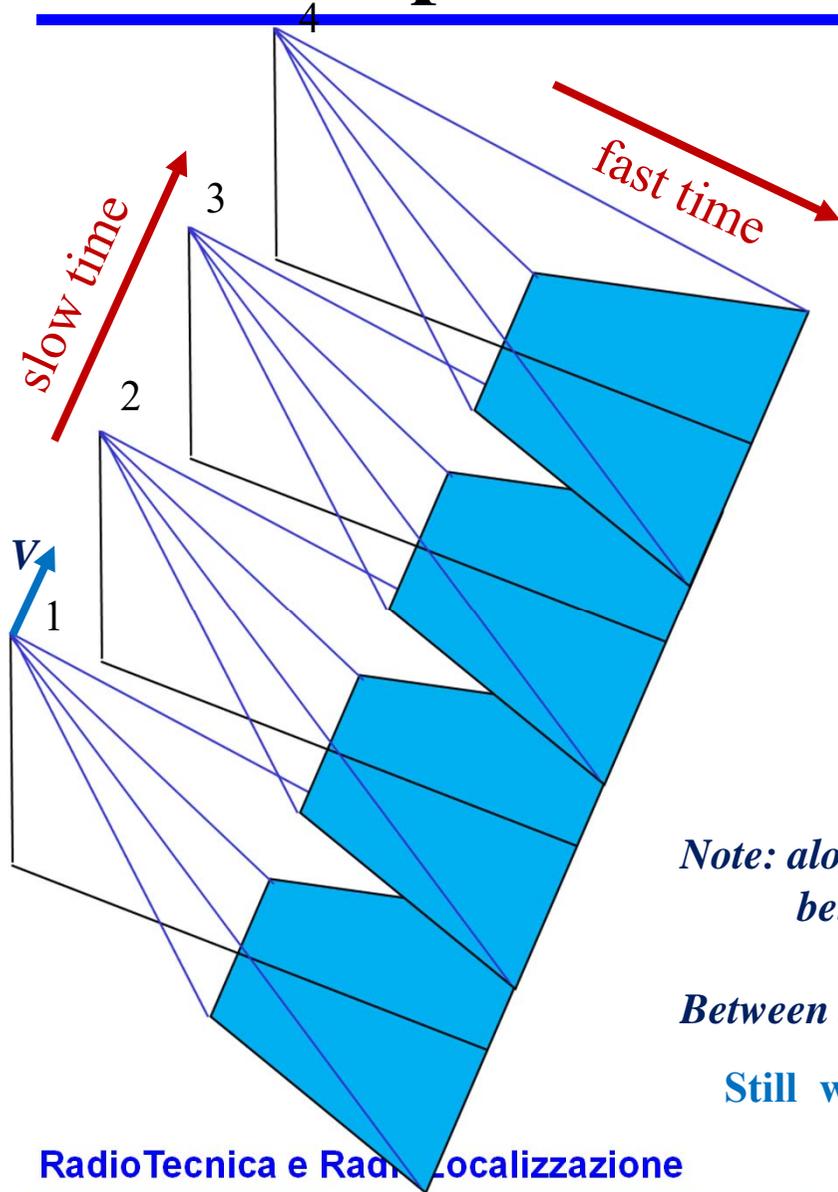
Two problems:

- 1 pixel represents a ground patch of: 0.5 m × 344 m !!!!
- Vector collecting “fast time” samples: not a matrix – not an image !!!!



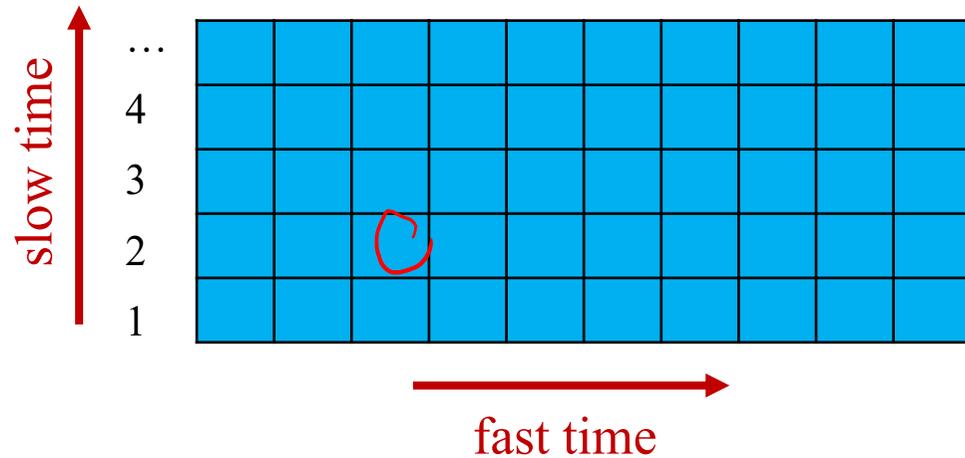
fast time

Real Aperture Radar



Exploit platform motion in “slow time”

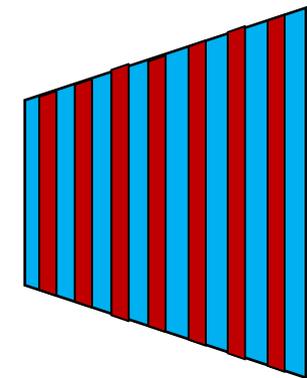
- pulses collected from different positions
- a matrix → an image !!!!



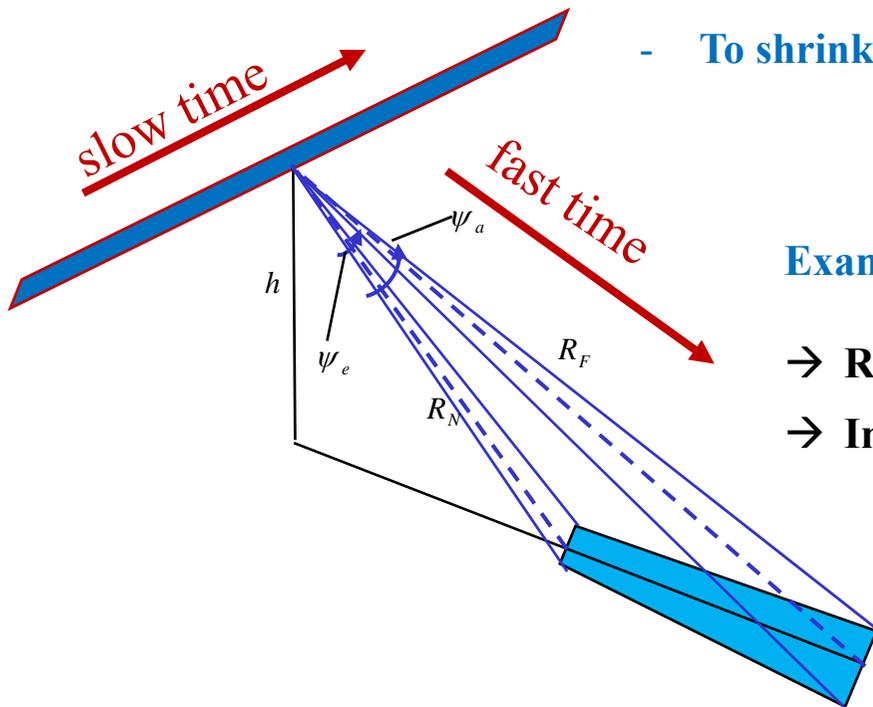
*Note: along slow time ... space = $V * \text{time}$
being V the platform velocity*

Between two pulses displacement of $V \text{ PRT}$

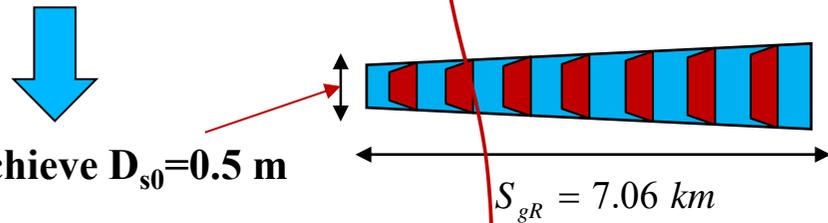
Still wide pixel along slow time !!!



Real Aperture Radar (II)



- To shrink resolution cell \rightarrow increase antenna length d_a



Example: to achieve $D_{s0} = 0.5$ m

\rightarrow Reduce beamwidth $344/0.5 = 688$ times!

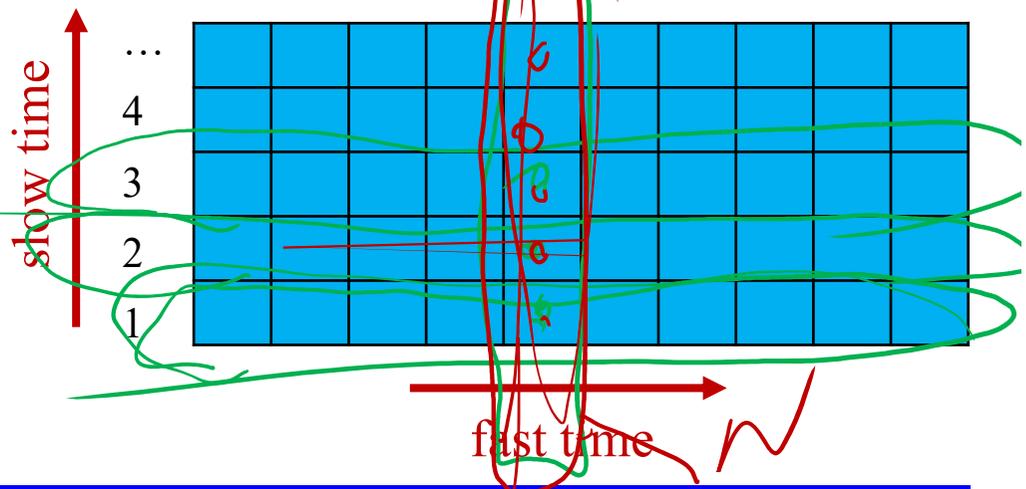
\rightarrow Increase antenna length d_a 688 times:

$d_a = 1.8 * 688 = 1238.4$ m !!!!!!! IMPOSSIBLE!!!

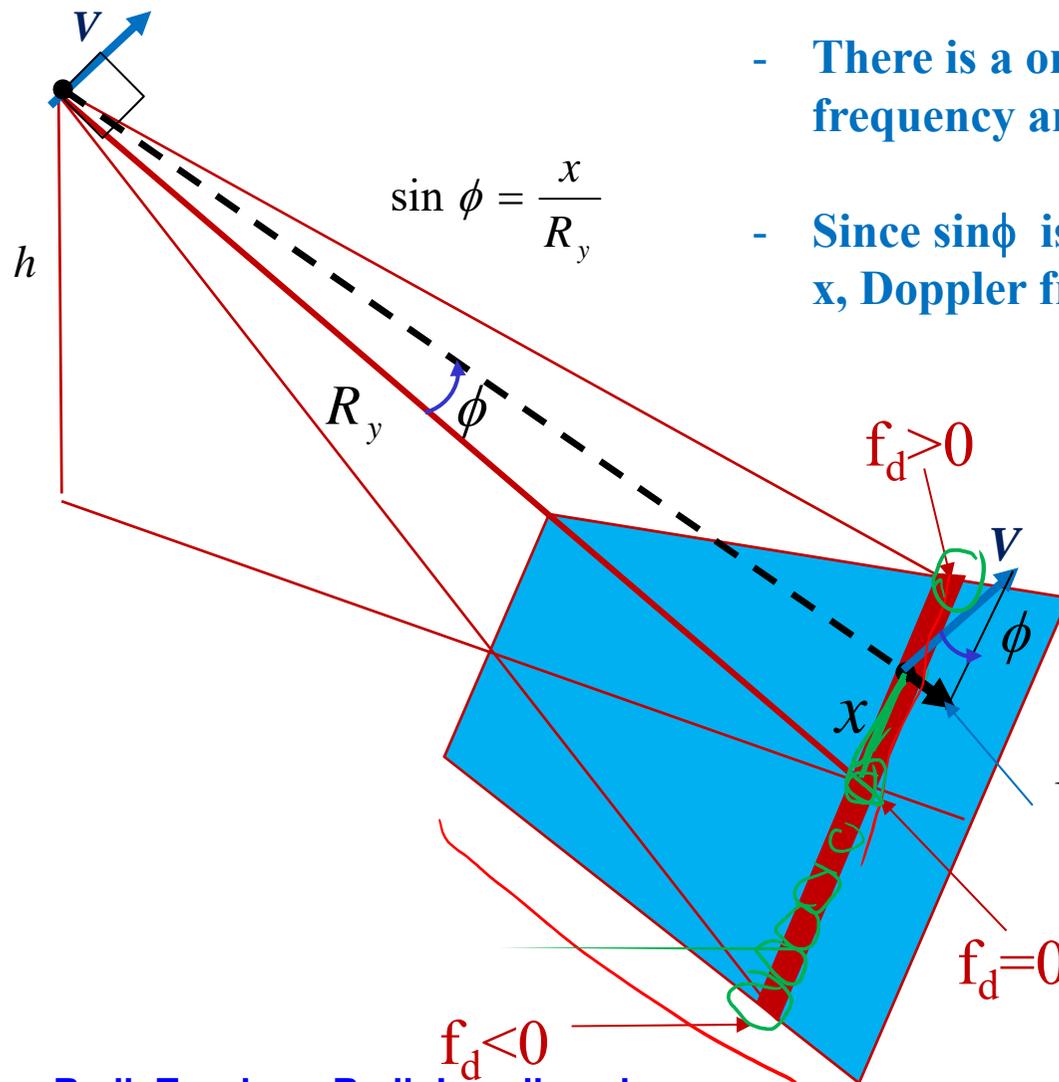
Using platform motion:

\rightarrow a matrix \rightarrow an image !!!!

with 0.5 m \times 0.5 m ground resolution



Angle-Doppler frequency relationship



$$\sin \phi = \frac{x}{R_y}$$

- There is a one-to-one relationship between Doppler frequency and \sin of angle ϕ
- Since $\sin \phi$ is linear with along-track displacement x , Doppler frequency is linear with x

$$f_d = \frac{2}{\lambda} v_r = \frac{2}{\lambda} V \sin \phi$$

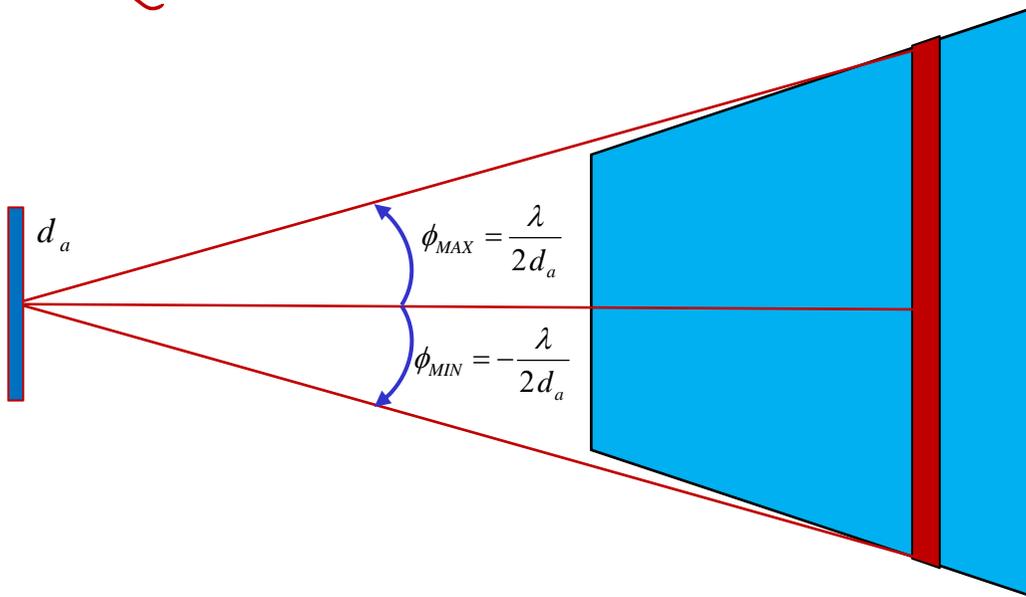
$$v_r = V \sin \phi$$

$$f_d \cong \frac{2V}{\lambda R_y} x$$

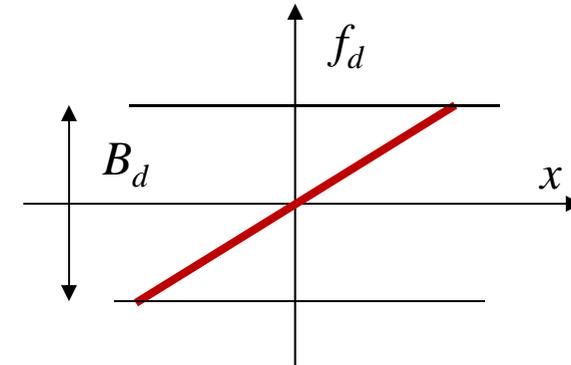
$$\Delta f_d = \frac{2V}{\lambda R_y} \Delta x$$

Doppler frequency bandwidth

$$f_{d \text{ MAX}} = \frac{2}{\lambda} V \sin \phi_{\text{MAX}} = \frac{2}{\lambda} V \sin\left(\frac{\lambda}{2d_a}\right) \cong \frac{2}{\lambda} V \frac{\lambda}{2d_a} = \frac{V}{d_a}$$

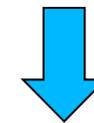


$$f_{d \text{ MIN}} = \frac{2}{\lambda} V \sin \phi_{\text{MIN}} \cong -\frac{V}{d_a}$$



Doppler frequency bandwidth:

$$B_d = f_{d \text{ MAX}} - f_{d \text{ MIN}} = \frac{2V}{d_a}$$



- Minimum PRF

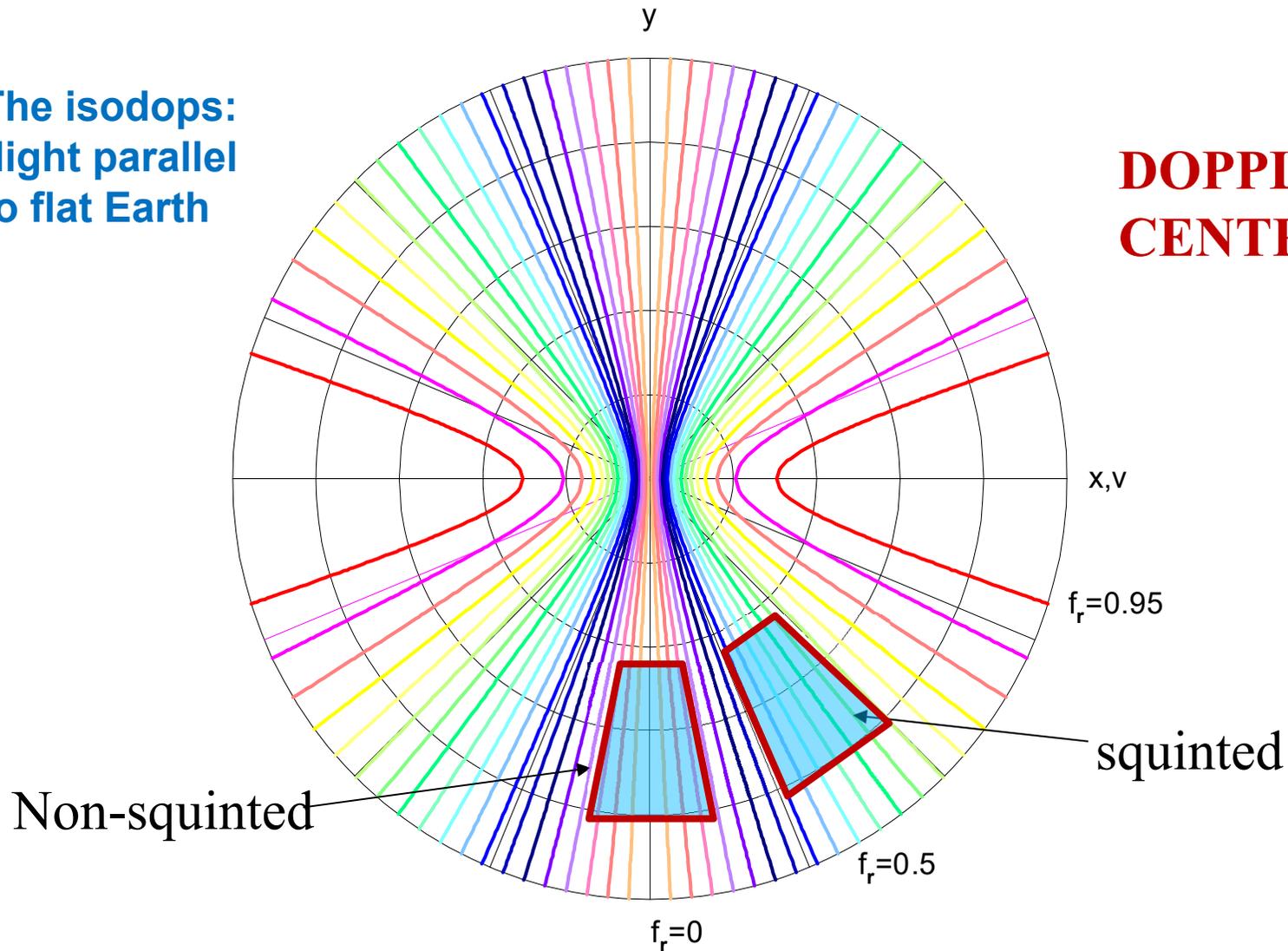
$$PRF \geq B_d = \frac{2V}{d_a}$$

$$PRF \leq \frac{d_a}{2V}$$

Frequency approach to SAR

The isodops:
flight parallel
to flat Earth

**DOPPLER
CENTROID**



Along-track resolution by Doppler

- Doppler frequency resolution (*Fourier Transform*)

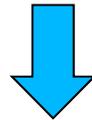
$$\Delta f_d = \frac{1}{T_{oss}} = \frac{1}{N \cdot PRT} = \frac{PRF}{N}$$

$$\Delta f_d = \frac{1}{T_{oss}} = \frac{2}{\lambda} V \delta \sin \phi$$

$$\delta \sin \phi = \frac{\lambda}{2V} \Delta f_d = \frac{\lambda}{2V} \frac{1}{T_{oss}} = \frac{\lambda}{2V} \frac{PRF}{N}$$

$$\Delta f_d = \frac{1}{T_{oss}} = \frac{2V}{\lambda R_y} \delta x$$

$$\delta x = \frac{\lambda R_y}{2V} \Delta f_d = \frac{\lambda R_y}{2V} \frac{1}{T_{oss}} = \frac{\lambda R_y PRF}{2V N}$$

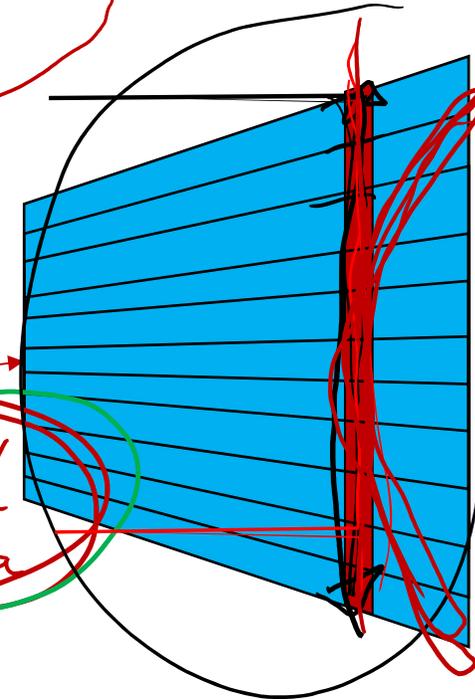


- N pulses at min PRF: FFT provides N Doppler filters

$$\delta \sin \phi \geq \frac{\lambda}{2V} \frac{2V}{N d_a} = \frac{1}{N} \frac{\lambda}{d_a} = \frac{\psi_a}{N}$$

$$\delta x \geq \frac{\lambda R_y}{2V} \frac{2V}{N d_a} = \frac{1}{N} \left(\frac{\lambda}{d_a} R_y \right) = \frac{D_{sy}}{N}$$

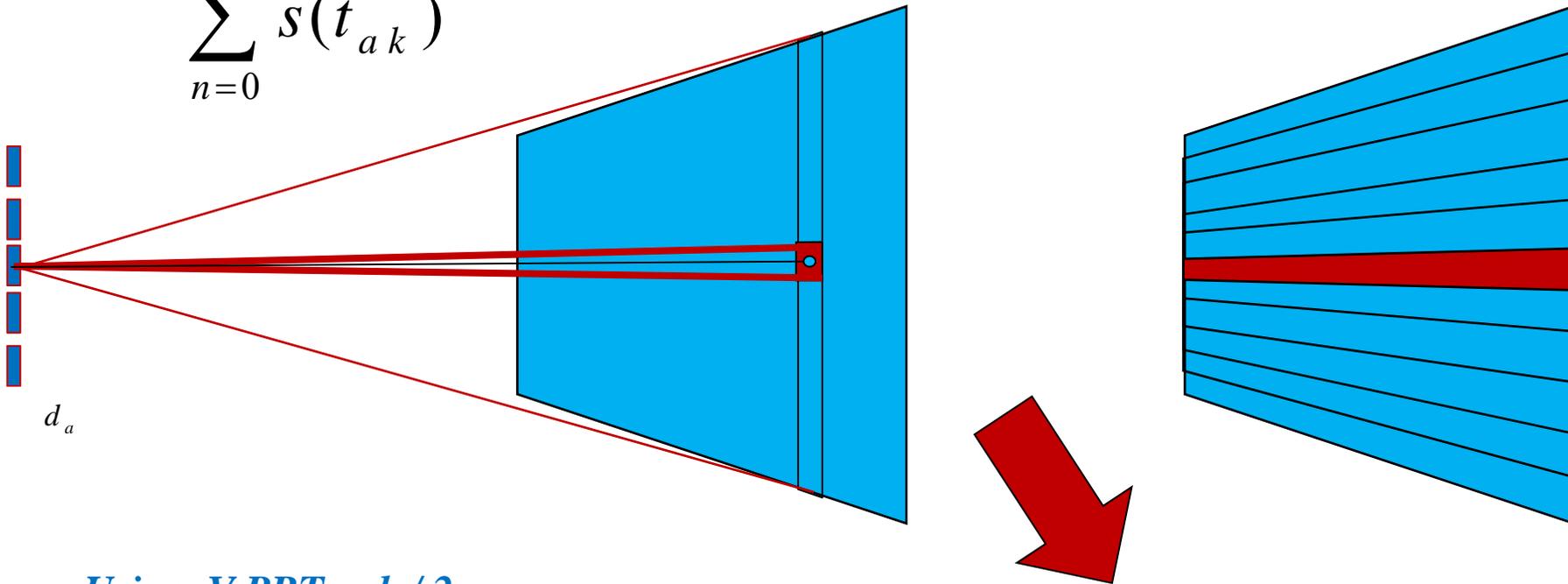
$$PRF \Rightarrow \frac{2V}{d_a}$$



synthetic antenna principle

- By exploiting platform motion emulate “synthetic antenna array”

$$\sum_{n=0}^{N-1} s(t_{ak})$$



-Using $V PRT = d_a / 2$:

$$\frac{2V PRT}{\lambda} \delta \sin \phi = \frac{1}{N} \rightarrow \delta \sin \phi = \frac{\lambda}{2V \cdot PRT} \frac{1}{N} = \frac{\lambda}{d_a} \frac{1}{N} = 2 \frac{\psi_a}{N}$$

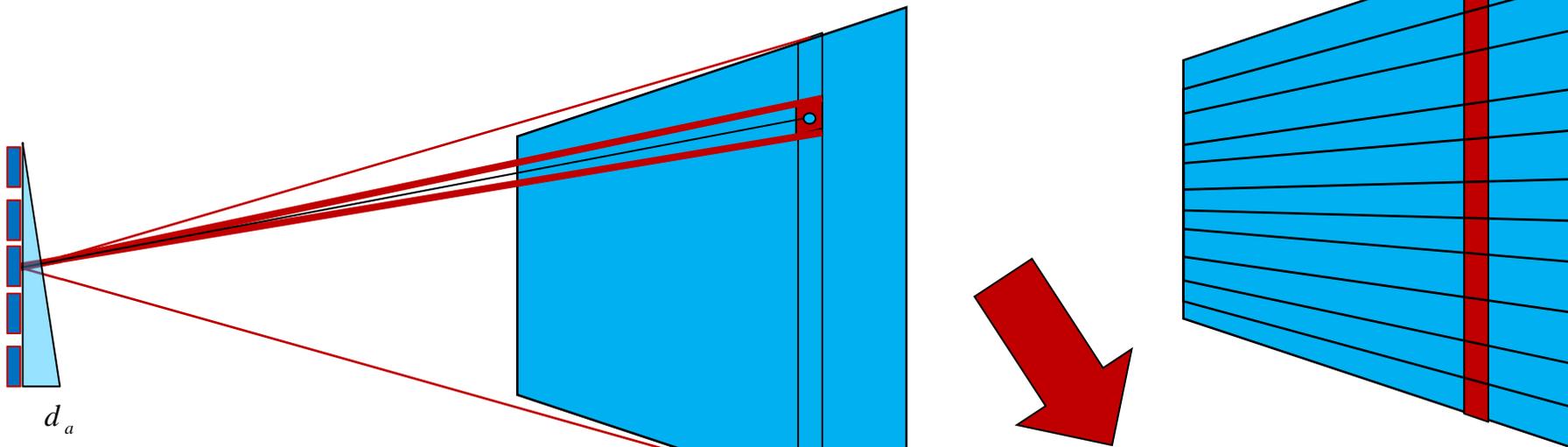
Synthetic antenna beam N times narrower than real antenna beam

synthetic antenna principle (II)

- By exploiting platform motion emulate “synthetic antenna array”

-To steer in direction ϕ , add all returns of the N pulses after compensating a linear phase term $\Delta\phi = 2\pi k \frac{d}{\lambda} \sin\phi$ **both in TX and in RX** (twice as in standard array):

$$\sum_{n=0}^{N-1} s(t_{ak}) e^{-j2\left[2\pi \frac{nd}{\lambda} \sin\phi\right]} = \sum_{n=0}^{N-1} s(n \text{ PRT}) e^{-j4\pi \frac{nV \cdot \text{PRT}}{\lambda} \sin\phi} = \text{FFT} \left\{ s(n \text{ PRT}) \right\}_{k=\frac{2V \cdot \text{PRT}}{\lambda} \sin\phi}$$

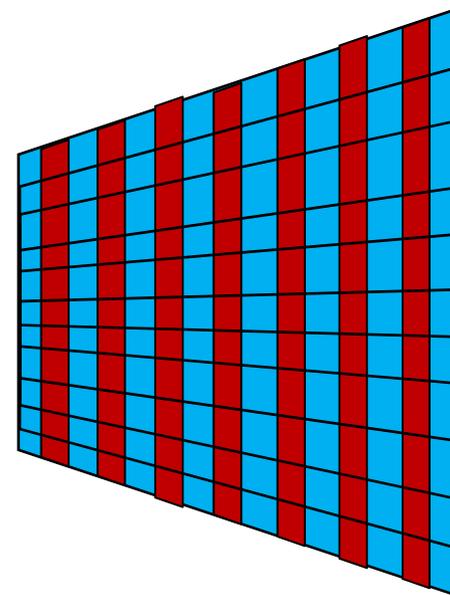
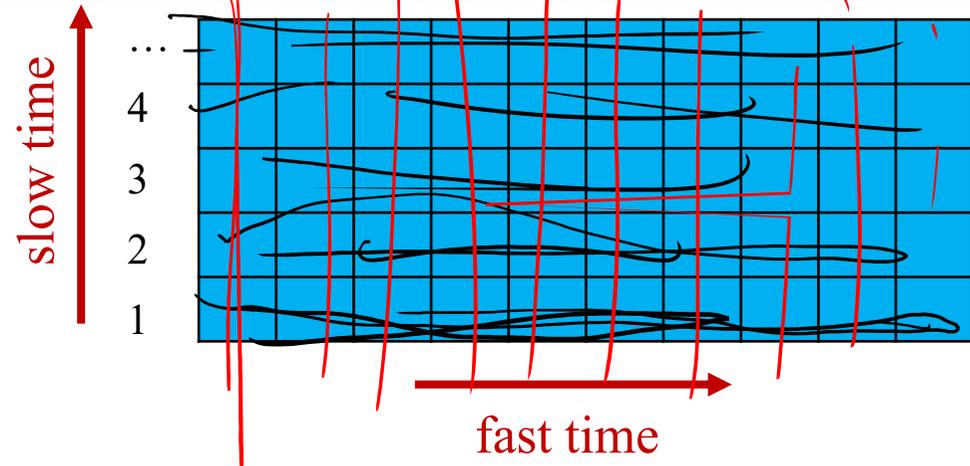
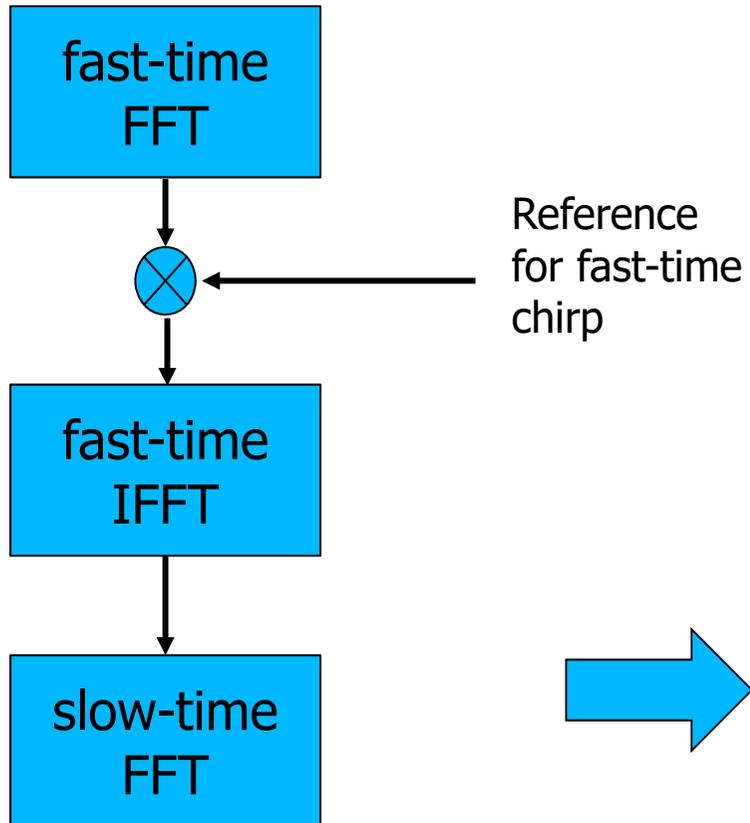


-Using $V \text{ PRT} = d_a / 2$:

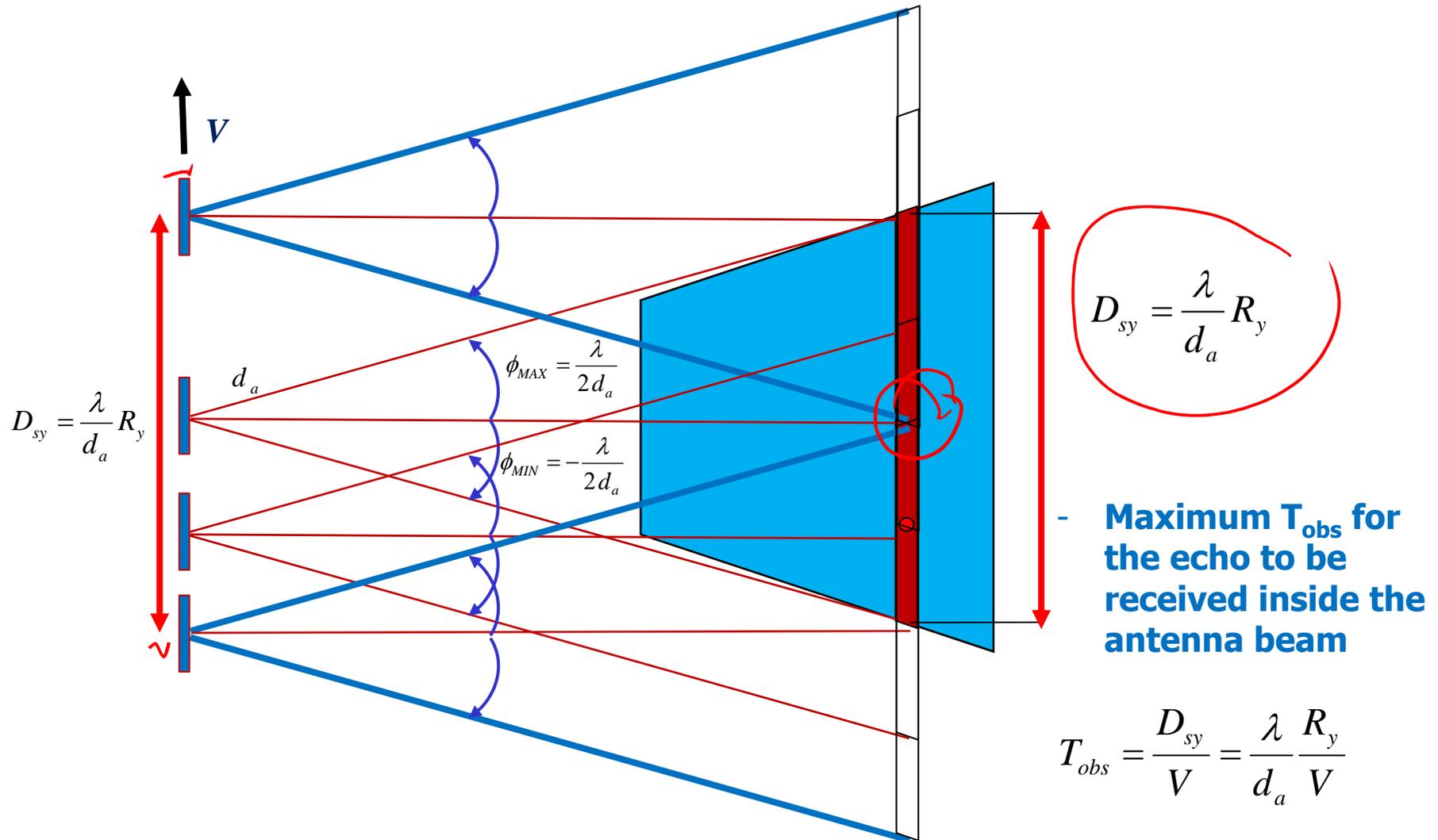
$$\frac{2V \text{ PRT}}{\lambda} \delta \sin\phi = \frac{1}{N} \rightarrow \delta \sin\phi = \frac{\lambda}{2V \cdot \text{PRT}} \frac{1}{N} = \frac{\lambda}{d_a} \frac{1}{N} = 2 \frac{\psi_a}{N}$$

Synthetic antenna beam N times narrower than real antenna beam

Unfocused SAR Processing scheme



Maximum observation time for point target



SAR azimuth resolution

To exploit long T_{oss} we can think in terms of:

- **Narrow Doppler filter at zero Doppler using the whole T_{oss}**

may

$$T_{oss} = \frac{D_{sy}}{V} = \frac{\lambda R_y}{d_a V}$$

$$\delta x = \frac{\lambda R_y}{2V} \Delta f_d = \frac{\lambda R_y}{2V} \frac{1}{T_{oss}} = \frac{\lambda R_y}{2V} \frac{1}{\frac{D_{sy}}{V}} = \frac{\lambda R_y}{2V} \frac{1}{\frac{\lambda R_y}{d_a V}} = \frac{d_a}{2}$$

- To achieve high resolution -> Small-sized ANTENNA appears better !

Limit to Doppler frequency resolution

Longer T_{oss} = longer pulse sequence \rightarrow Higher Doppler frequency resolution

$$\Delta f_d = \frac{1}{T_{oss}} = \frac{PRF}{N}$$

- This all applies as long as the platform motion does not force motion of point on ground out of the Doppler filter

$$\left\{ \begin{aligned} \delta \sin \phi &= \frac{\lambda}{2V} \frac{1}{T_{oss}} \\ \delta x &= \frac{\lambda R_y}{2V} \frac{1}{T_{oss}} \end{aligned} \right.$$

$$V T_{oss} \leq \delta x = \frac{\lambda R_y}{2V} \frac{1}{T_{oss}}$$

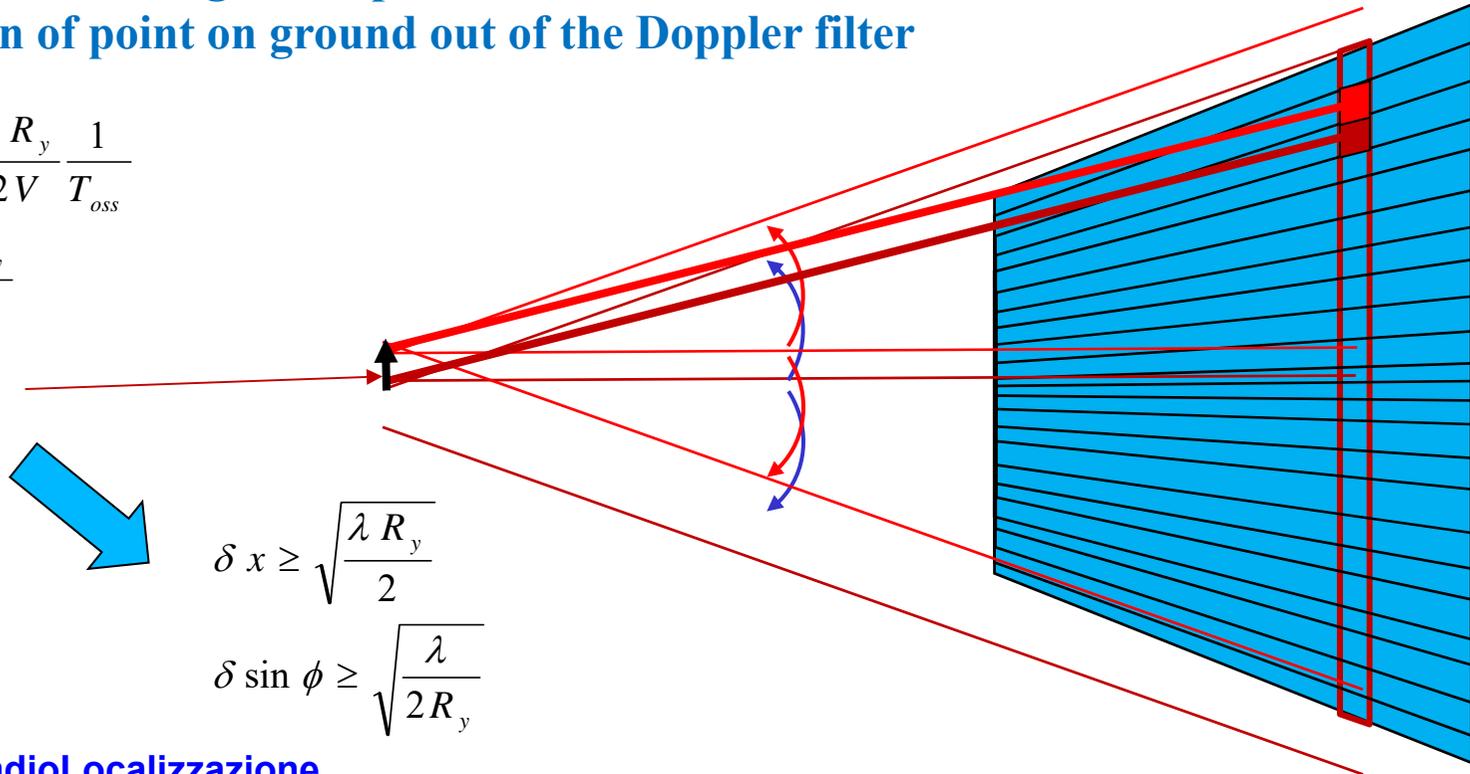
$$(V T_{oss})^2 \leq \frac{\lambda R_y}{2}$$

$$V T_{oss} \leq \sqrt{\frac{\lambda R_y}{2}}$$

Maximum resolution:

$$\delta x \geq \sqrt{\frac{\lambda R_y}{2}}$$

$$\delta \sin \phi \geq \sqrt{\frac{\lambda}{2R_y}}$$



Max unfocused SAR resolution

Longer T_{oss} = longer pulse sequence \rightarrow Higher Doppler frequency resolution

$$\Delta f_d = \frac{1}{T_{obs}} = \frac{PRF}{N}$$

$$V T_{oss} \leq \sqrt{\frac{\lambda R_y}{2}} = \begin{cases} \sqrt{\lambda R_N / 2} = 16.45 \text{ m} \\ \sqrt{\lambda R_0 / 2} = 17.61 \text{ m} \\ \sqrt{\lambda R_F / 2} = 19.13 \text{ m} \end{cases}$$

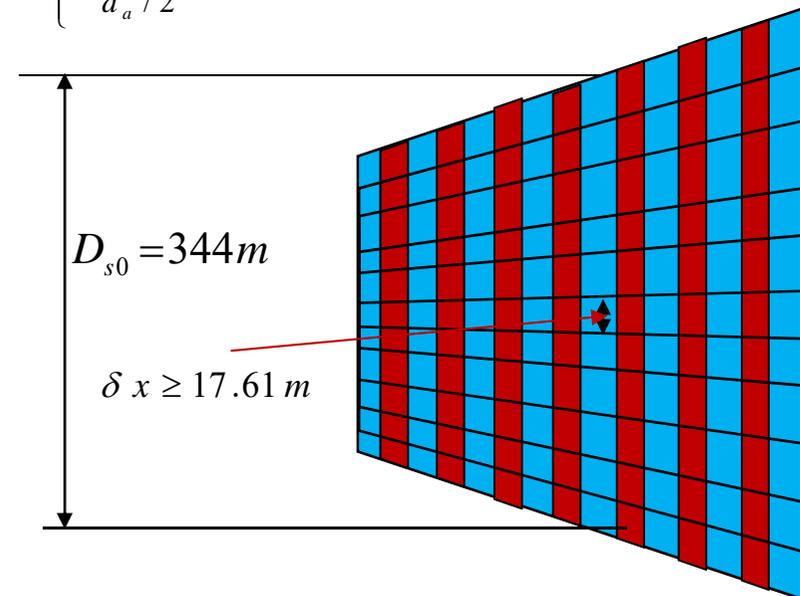
Maximum resolution:

$$\delta x \geq \sqrt{\frac{\lambda R_y}{2}} = \begin{cases} \sqrt{\lambda R_N / 2} = 16.45 \text{ m} \\ \sqrt{\lambda R_0 / 2} = 17.61 \text{ m} \\ \sqrt{\lambda R_F / 2} = 19.13 \text{ m} \end{cases}$$

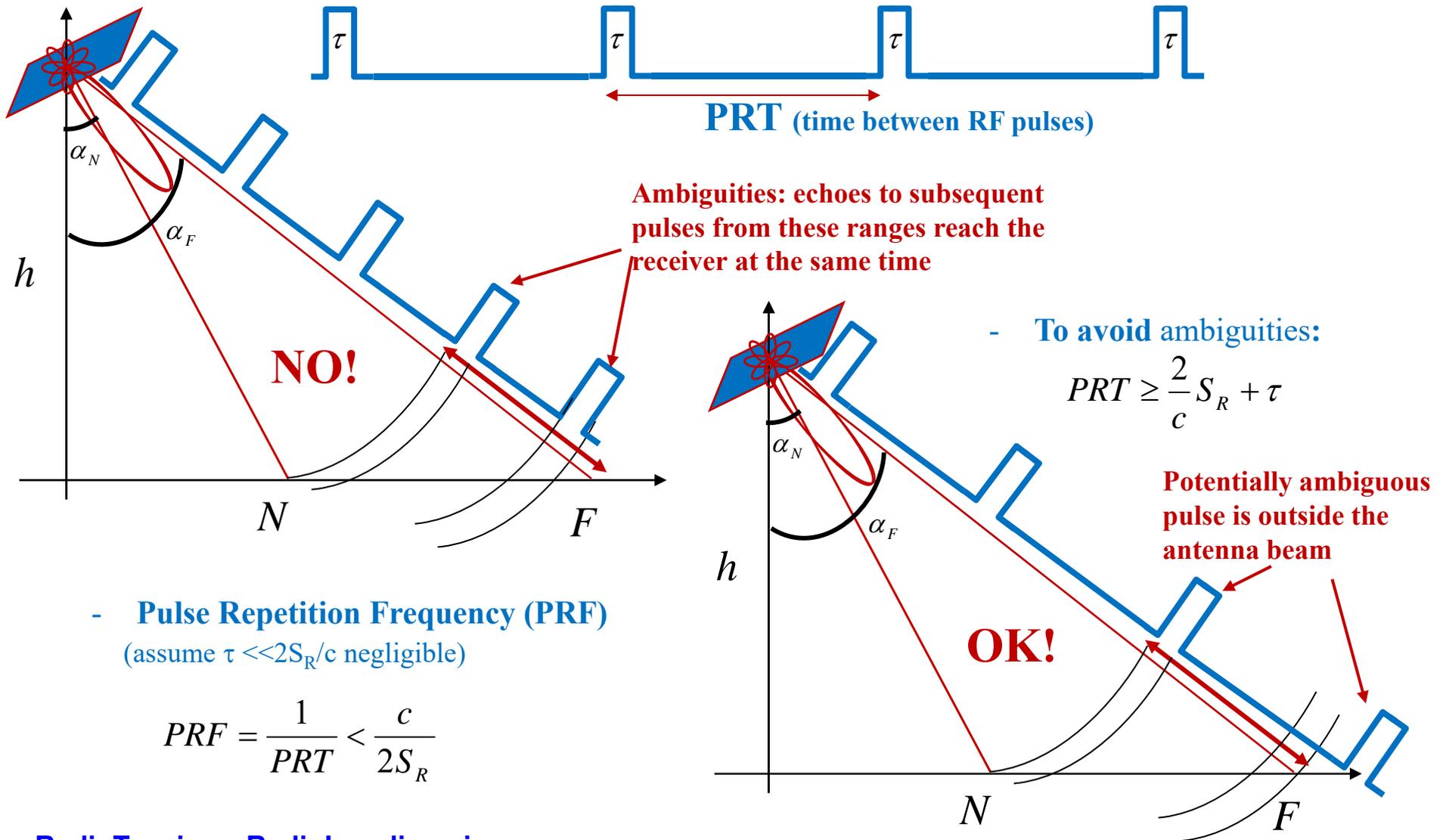
$$\delta \sin \phi \geq \sqrt{\frac{\lambda}{2 R_y}} = \begin{cases} \sqrt{\frac{\lambda}{2 R_N}} \rightarrow \phi \cong 0.0540^\circ \\ \sqrt{\frac{\lambda}{2 R_F}} \rightarrow \phi \cong 0.0464^\circ \end{cases}$$

$$N = T_{obs} PRF = T_{obs} B_d = T_{obs} \frac{2V}{d_a} \leq \frac{2}{d_a} \sqrt{\frac{\lambda R_y}{2}}$$

$$N \leq \begin{cases} \frac{\sqrt{\lambda R_N / 2}}{d_a / 2} = 18.28 \\ \frac{\sqrt{\lambda R_0 / 2}}{d_a / 2} = 19.56 \\ \frac{\sqrt{\lambda R_F / 2}}{d_a / 2} = 21.25 \end{cases}$$



Range ambiguities



- Pulse Repetition Frequency (PRF)
(assume $\tau \ll 2S_R/c$ negligible)

$$PRF = \frac{1}{PRT} < \frac{c}{2S_R}$$

Fundamental limitation of SAR

Avoidance of Range Ambiguities: $1/PRF > 2 S_R/c$

Avoidance of Azimuth Ambiguities: $PRF > 2v/\lambda * \text{Antenna beamwidth AZ}$

Range Swath: $S_R = \psi_e R_o / \cos \alpha = \lambda/d_e R_o / \cos \alpha$
 Antenna beamwidth AZ: $\psi_a = \lambda/d_a$

$$\frac{2 v \lambda}{\lambda d_a} < PRF < \frac{c}{2} \frac{d_e \cos \alpha}{\lambda R_o}$$



$$\frac{2 v \lambda}{\lambda d_a} < \frac{c}{2} \frac{d_e \cos \alpha}{\lambda R_o}$$



$$\frac{S_R}{d_a/2} < \frac{c}{2v}$$



$$d_e d_a > \frac{4 v \lambda R_o}{c \cos \alpha}$$