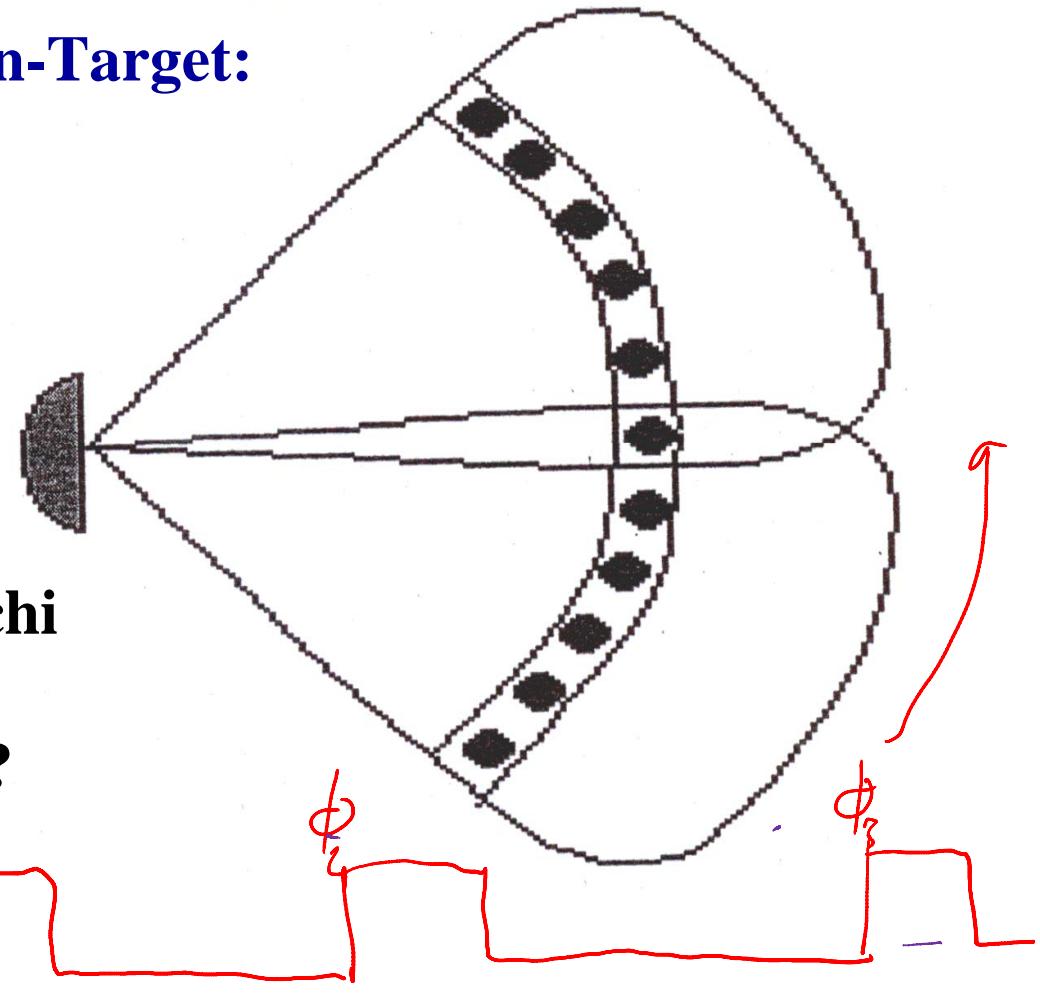
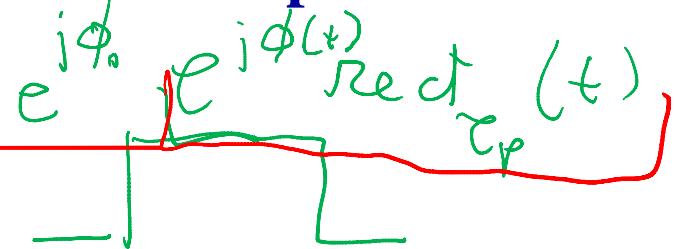

Integrazione non coerente per bersagli fissi

Pierfrancesco Lombardo

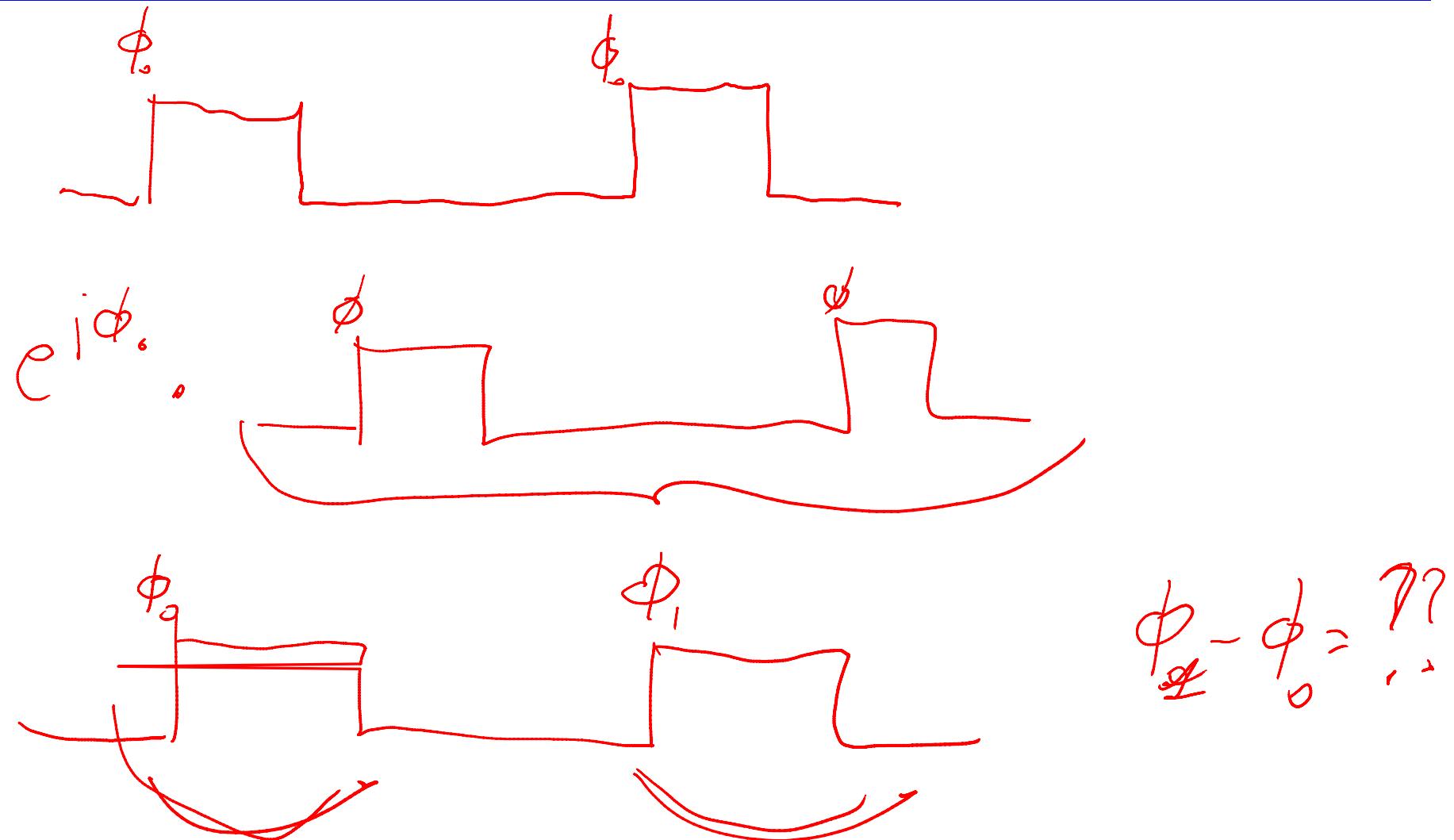
Integrazione di impulsi

– N impulsi nel Time-on-Target:



Come sfruttare gli N echi
per massimizzare la
capacità di rivelazione?





$$P(\underline{r} \leq \underline{r}) = \frac{1}{(\pi \sigma^2)^N} e^{-\frac{1}{\sigma^2} [\underline{r} - \underline{a}]^T [\underline{r} - \underline{a}]}$$

$$\underline{s} = \begin{bmatrix} s_0 \\ s_1 \\ \vdots \\ s_M \end{bmatrix}$$

$$\underline{r} = \begin{bmatrix} r_0 \\ r_1 \\ r_2 \\ r_3 \\ \vdots \end{bmatrix} = \underline{a} + \underline{s}_0 + \underline{n}$$

$$P(\underline{r} \leq \underline{r} | H_1) = \frac{1}{(\pi \sigma_n^2)^N} e^{-\frac{1}{\sigma_n^2} \|\underline{r} - \underline{a}\|^2} *$$

$$P(\underline{r} \leq \underline{r} | H_0) = \frac{1}{(\pi \sigma_n^2)^N} e^{-\frac{1}{\sigma_n^2} \|\underline{r}\|^2} *$$

Sistemi Radar

$$\Delta = \frac{P(r/\mu_2)}{P(r/\mu_1)} = e^{-\frac{1}{6n^2} \left\{ \|r - a_S\|^2 - \|r\|^2 \right\}} > T$$

$$\lambda \approx \ln \Delta = -\frac{1}{6n^2} \left\{ \|r - a_S\|^2 - \|r\|^2 \right\} > T$$

$$[r - a_S]^H [r - a_S] = r^H r - a^* S^H r + a^* a_S S^H S = \\ = \|r\|^2 - 2 \operatorname{Re}\{a^* r^H S\} + \|a\|^2 \|S\|^2$$

$$* \lambda = \frac{1}{6n^2} \|a\|^2 \|S\|^2 + \frac{2}{6n^2} \operatorname{Re}\{a^* r^H S\} *$$

Sistemi Radar

$$x = -\frac{|a|^2}{G_n^2} \|s\|^2 + \frac{2}{G_n^2} |a| |\underline{r^h s}| \cos(\angle a + \angle \underline{r^h s})$$

$$\begin{aligned} a \underline{r^h s} &= |a| e^{i \angle a} \cdot |\underline{r^h s}| e^{i \angle \underline{r^h s}} \\ \operatorname{Re}\{a \underline{r^h s}\} &= |a| |\underline{r^h s}| \operatorname{Re}\{e^{i [\angle a + \angle \underline{r^h s}]}\} \\ &= |a| |\underline{r^h s}| \cos(\angle a + \angle \underline{r^h s}) \end{aligned}$$

Rivelazione GLRT (I)

- Rapporto di verosimiglianza :

$$\lambda = \exp \left\{ \frac{2}{\sigma_n^2} |a| \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r} \right| \cos \left[-\angle a + \angle \mathbf{s}_0^H [t_0, f_0] \mathbf{r} \right] \right\} \exp \left\{ -\frac{|a|^2}{\sigma_n^2} \right\}$$

X

- Rapporto di verosimiglianza generalizzato, ottenuto massimizzando λ sulla ddp della fase di a :

$$\ln \lambda = \frac{2}{\sigma_n^2} |a| \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r} \right| \cos \left[-\angle a + \angle \mathbf{s}_0^H [t_0, f_0] \mathbf{r} \right] - \frac{|a|^2}{\sigma_n^2} / > T ?$$

$$\star \max_{\angle a} \{\ln \lambda\} = \frac{2}{\sigma_n^2} |a| \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r} \right| - \frac{|a|^2}{\sigma_n^2} > T$$

$$\frac{\partial}{\partial |a|} \left[\max_{\angle a} \{\ln \lambda\} \right] = \frac{2}{\sigma_n^2} \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r} \right| - \frac{2|a|}{\sigma_n^2} = 0$$

$$|a| = \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r} \right|$$

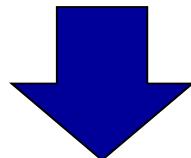
$$\max_{\angle a, |a|} \{\ln \lambda\} = \frac{2}{\sigma_n^2} \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r} \right|^2 - \frac{\left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r} \right|^2}{\sigma_n^2} = \frac{1}{\sigma_n^2} \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r} \right|^2 > T$$

Rivelazione GLRT (II)

– Rapporto di verosimiglianza :

$$\lambda = \exp\left\{ \frac{2}{\sigma_n^2} |a| |\mathbf{s}_0^H[t_0, f_0] \mathbf{r}| \cos[-\angle a + \angle \mathbf{s}_0^H[t_0, f_0] \mathbf{r}] \right\} \exp\left\{ -\frac{|a|^2}{\sigma_n^2} \right\}$$

$$\max_{\angle a, |a|} \{\ln \lambda\} = \frac{1}{\sigma_n^2} |\mathbf{s}_0^H[t_0, f_0] \mathbf{r}|^2$$



$$\Lambda > T_d \quad \Leftrightarrow \quad |\mathbf{s}_0^H[t_0, f_0] \mathbf{r}| > T$$

Rivelazione GLRT per sequenza

- Rapporto di verosimiglianza per sequenza di N impulsi :**

- Se la fase delle riflessioni dal bersaglio è aleatoria e uniformemente distribuita, gli echi di ritorno a ciascun impulso inviato sono statisticamente indipendenti.
- Dunque la DDP della sequenza è ottenuta come prodotto delle DDP relative agli echi ai singoli impulsi.
- Di conseguenza, anche la funzione di verosimiglianza per la sequenza risulta pari al prodotto delle verosimiglianze per i singoli impulsi
- La massimizzazione del logaritmo della DDP congiunta rispetto alle ampiezze complesse degli echi bersagli, porta alla somma delle verosimiglianze generalizzate logaritmiche

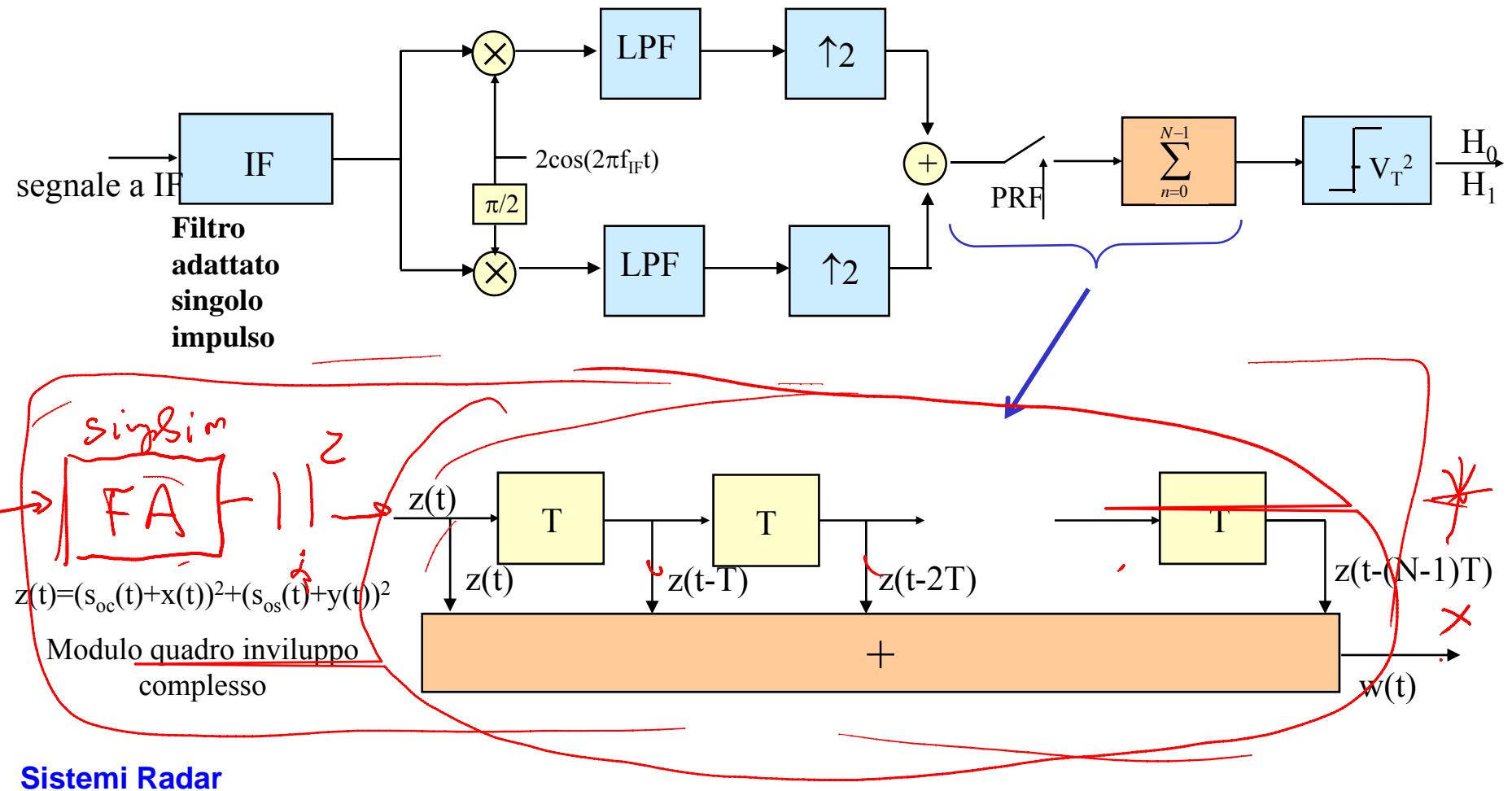
$$\lambda_{seq} = \prod_{n=0}^{N-1} \lambda_n = \prod_{n=0}^{N-1} \exp \left\{ \frac{2}{\sigma_n^2} |a_n| \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r}_n \right| \cos \left[-\angle a_n + \angle \mathbf{s}_0^H [t_0, f_0] \mathbf{r}_n \right] \exp \left\{ -\frac{|a_n|^2}{\sigma_n^2} \right\} \right\}$$

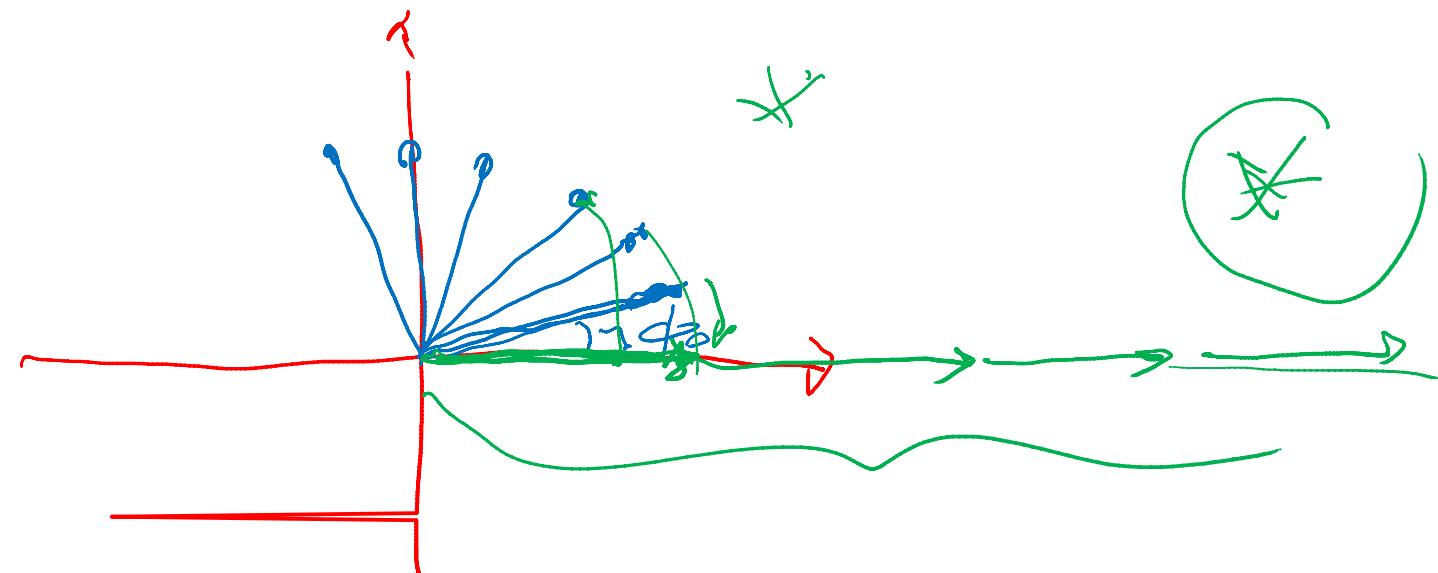
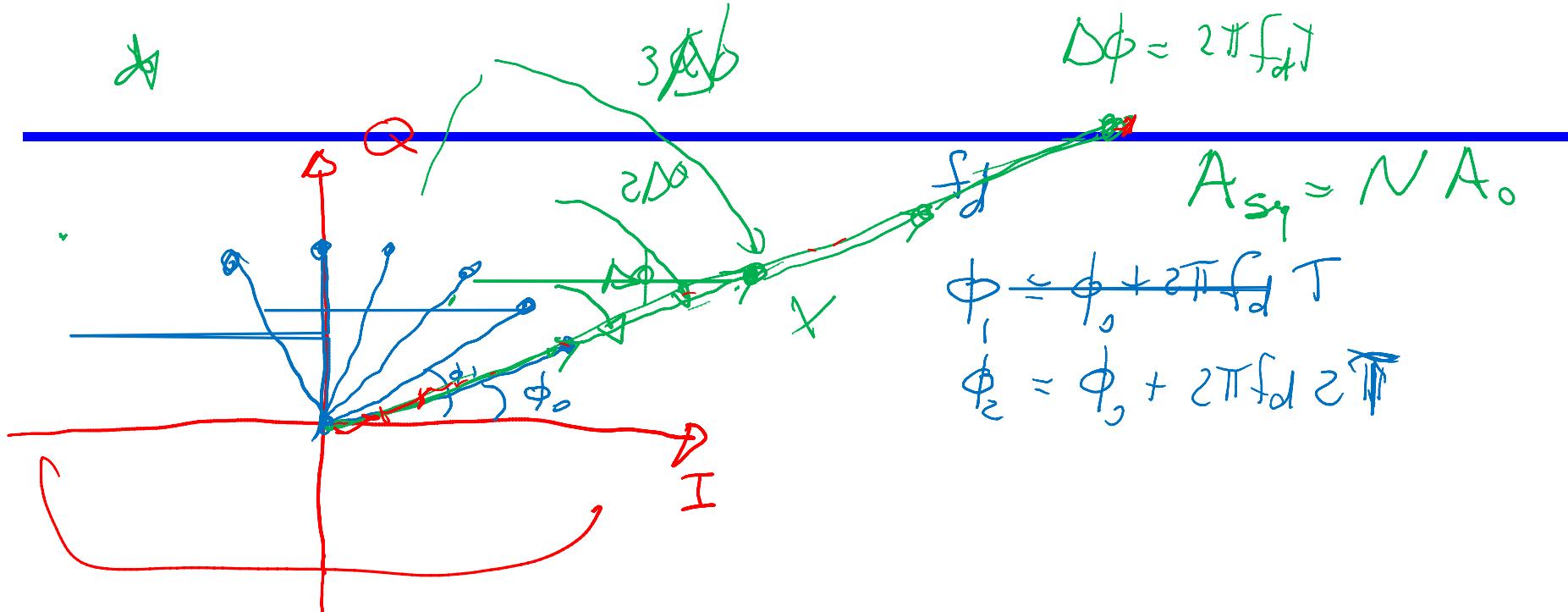
$$\max_{\angle a_0, |a_0|, \dots, \angle a_{N-1}, |a_{N-1}|} \{ \ln \lambda_{seq} \} = \sum_{n=0}^{N-1} \max_{\angle a_n, |a_n|} \{ \ln \lambda_n \} = \frac{1}{\sigma_n^2} \sum_{n=0}^{N-1} \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r}_n \right|^2$$

$$\Lambda > T_d \quad \Leftrightarrow \quad \sum_{n=0}^{N-1} \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r}_n \right|^2 > T$$

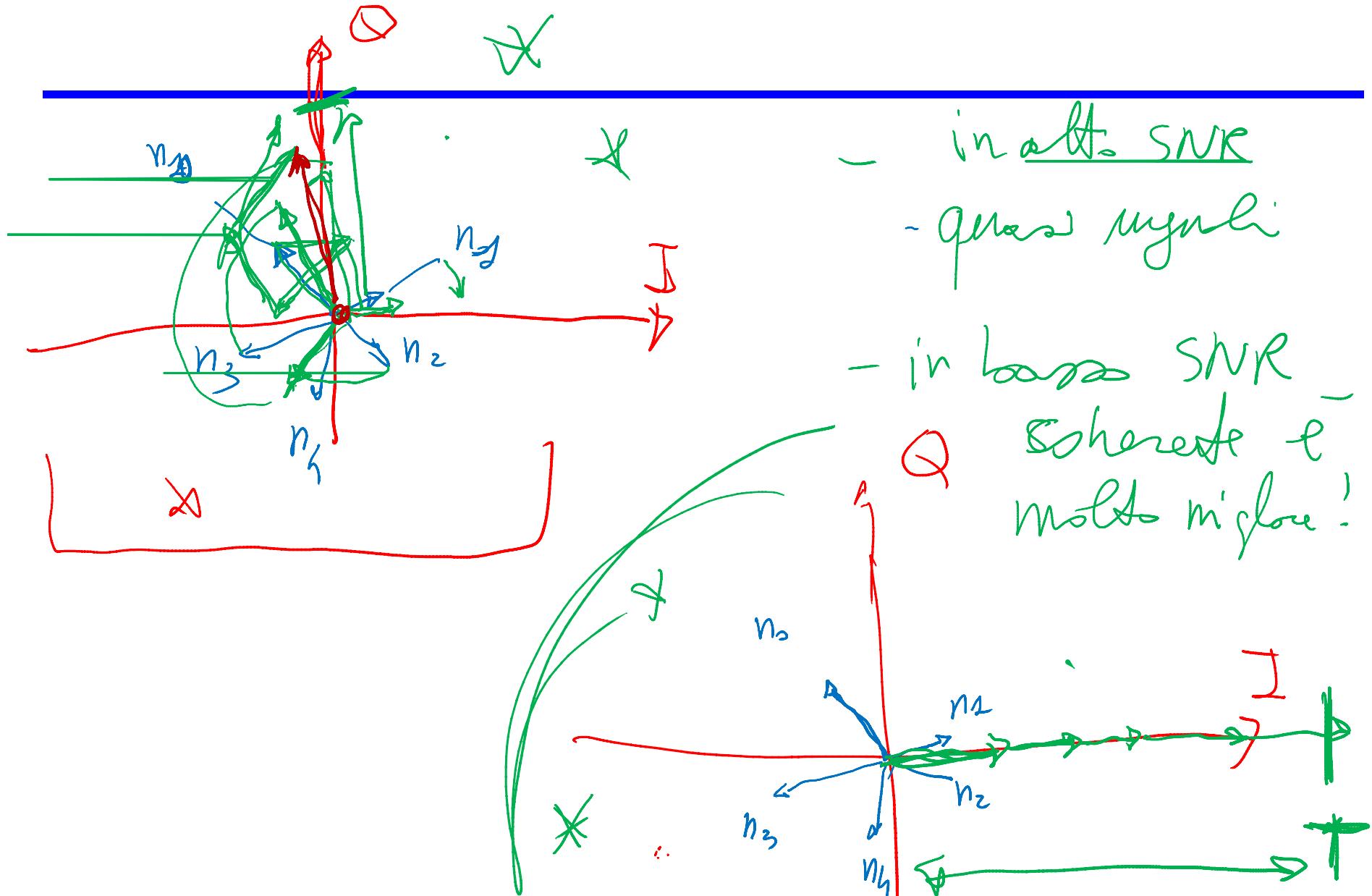
Integrazione incoerente quadratica (I)

- Modulo quadro dell'inviluppo complesso;
- Somma dei moduli quadri degli inviluppi complessi associati agli N impulsi

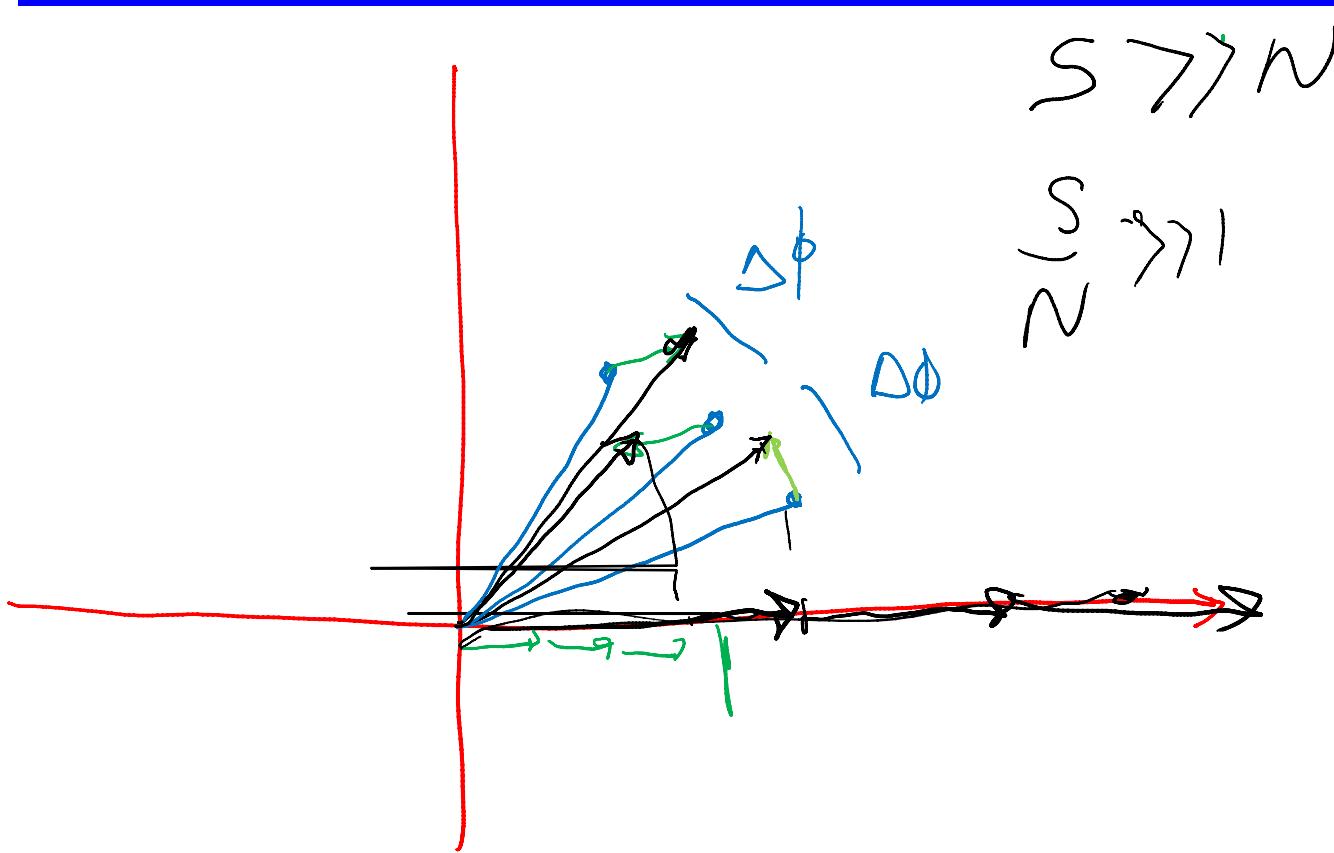




Sistemi Radar

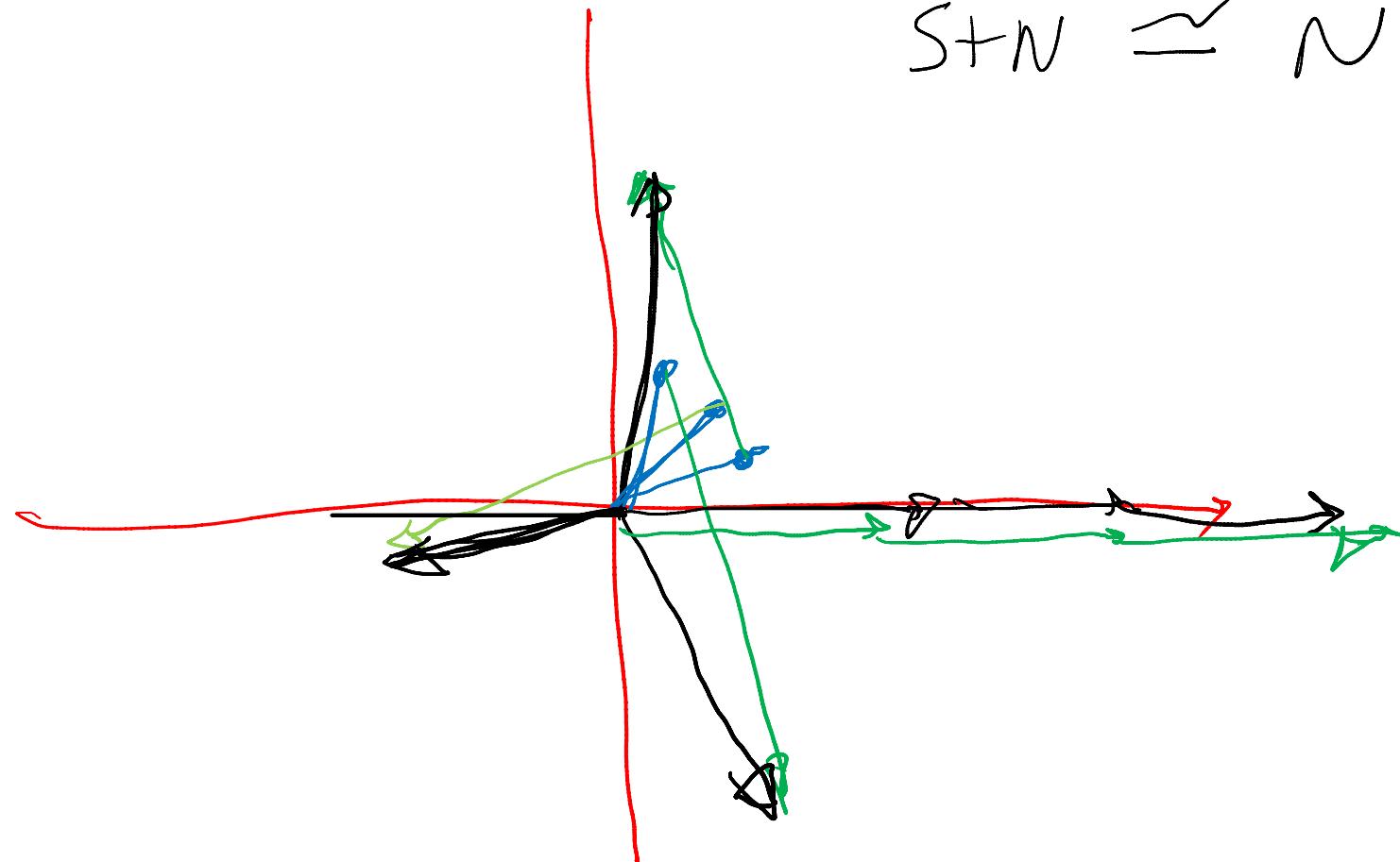


Sistemi Radar



Sistemi Radar

$$S+N \simeq N$$



Sistemi Radar

Rivelatore coerente

$P_d = e^{-\frac{T^2}{\sigma^2}}$

$P_d = \text{fus. d' Mercurio} (P_a, SNR)$

$P_a = 10^{-6}$ $P_d = 0,9 \Rightarrow SNR = 13 dB \Rightarrow Z_0$

$SNR_0 = SNR_{Si} = \frac{2J}{N}$

$|SNR_0|_{dB} = |SNR_{Si}|_{dB} = 13 + 10 \log_{10} N^*$

Sistemi Radar

Rivelazione non coerente (NCI) (I)

– Test:

$$\Lambda > T_d$$

$$\Leftrightarrow \left| \mathbf{s}_0^H[t_0, f_0] \mathbf{r}_0 \right|^2 + \left| \mathbf{s}_0^H[t_0, f_0] \mathbf{r}_1 \right|^2 + \cdots + \left| \mathbf{s}_0^H[t_0, f_0] \mathbf{r}_{N-1} \right|^2 > T^2$$

$$\xi = z^2 - \sum_{n=0}^{N-1} z_n^2 = \sum_{n=0}^{N-1} \xi_n = \sum_{n=0}^{N-1} \left| \mathbf{s}_0^H[t_0, f_0] \mathbf{r}_n \right|^2 > T^2$$

Sotto l'ipotesi H_0 ($a=0$) ho un Falso Allarme

$$p_{\mathbf{r}_n}(\mathbf{r}_n | H_0) = \frac{1}{\pi^N \sigma_n^{2N}} \exp \left\{ -\frac{1}{\sigma_n^2} |\mathbf{r}_n|^2 \right\} \quad \rightarrow$$

$$p_{\xi_n}(\xi_n | H_0) = \frac{1}{\sigma_d^2} \exp \left\{ -\frac{1}{\sigma_d^2} \xi_n \right\}$$

$$p(\xi | H_0) = p_{\xi_0}(\xi | H_0) * p_{\xi_1}(\xi | H_0) * \cdots * p_{\xi_{N-1}}(\xi | H_0)$$

$$p_{\xi}(\xi | H_0) = \frac{1}{(N-1)! \sigma_d^{2N}} \xi^{N-1} e^{-\frac{\xi}{\sigma_d^2}}$$

Gamma

Sistemi Radar

$$P_{fa} = \Pr \{ \xi > T^2 | H_0 \} = \int_{T^2}^{\infty} p_{\xi}(\xi | H_0) d\xi = \sum_{n=0}^{N-1} \frac{1}{n!} \left(\frac{T^2}{\sigma_d^2} \right)^n e^{-\frac{T^2}{\sigma_d^2}}$$

Richiamo DDP NCI (I)



$$p_{\xi_n}(\xi_n | H_0) = \frac{1}{\sigma_d^2} \exp \left\{ -\frac{1}{\sigma_d^2} \xi_n \right\}$$

$$C_{\xi_n}(\omega | H_0) = \int_0^\infty \frac{1}{\sigma_d^2} \exp \left\{ -\frac{1}{\sigma_d^2} \xi_n \right\} e^{-j\omega \xi_n} d\xi_n = \frac{1}{\sigma_d^2} \frac{1}{\frac{1}{\sigma_d^2} + \omega} = \frac{1}{1 + \omega \sigma_d^2}$$

$$p(\xi | H_0) = p_{\xi_0}(\xi | H_0) * p_{\xi_1}(\xi | H_0) * \dots * p_{\xi_{N-1}}(\xi | H_0)$$

$$p_\xi(\xi | H_0) = \frac{1}{(N-1)!} \frac{\xi^{N-1}}{\sigma_d^{2N}} e^{-\frac{\xi}{\sigma_d^2}}$$

$$C_\xi(\omega | H_0) = \prod_{n=0}^{N-1} C_{\xi_n}(\omega | H_0) = C_{\xi_n}^N(\omega | H_0) = \frac{1}{(1 + \omega \sigma_d^2)^N}$$

$$P_{fa} = \Pr\{\xi > T^2 | H_0\} = \int_{T^2}^\infty p_\xi(\xi | H_0) dz$$

$$\int_x^{+\infty} \frac{1}{(N-1)!} t^{N-1} e^{-t} dt = \sum_{n=0}^{N-1} \frac{x^n}{n!} e^{-x}$$

per N intero > 0

$$P_{fa} = \int_{T^2}^\infty \frac{1}{(N-1)!} \frac{\xi^{N-1}}{\sigma_d^{2N}} e^{-\frac{\xi}{\sigma_d^2}} d\xi = \int_{T^2/\sigma_d^2}^\infty \frac{1}{(N-1)!} t^{N-1} e^{-t} dt = \sum_{n=0}^{N-1} \frac{1}{n!} \left(\frac{T^2}{\sigma_d^2} \right)^n e^{-\frac{T^2}{\sigma_d^2}}$$

Sistemi Radar

Rivelazione non coerente (NCI) (II)

– Test:

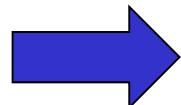
$$\Lambda > T_d \Leftrightarrow \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r}_0 \right|^2 + \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r}_1 \right|^2 + \cdots + \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r}_{N-1} \right|^2 > T^2$$

$$\xi = z^2 = \sum_{n=0}^{N-1} z_n^2 = \sum_{n=0}^{N-1} |\tilde{z}_n|^2 = \sum_{n=0}^{N-1} \xi_n = \sum_{n=0}^{N-1} \left| \mathbf{s}_0^H [t_0, f_0] \mathbf{r}_n \right|^2 > T^2$$

Sotto l'ipotesi H_1 ($a \neq 0$) ho una rivelazione

$$p(\tilde{z}|a; H_1) = \frac{1}{\pi \sigma_d^2} \exp \left\{ -\frac{1}{\sigma_d^2} |(\tilde{z} - a \cdot \chi(\tau, \nu))^2| \right\}$$

$$p_{\xi_n}(\xi_n | \sigma; H_1) = \frac{1}{\sigma_d^2} \exp \left\{ -\frac{\xi_n + \sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2} \right\} \cdot I_{\sigma} \left[\frac{2\sqrt{\xi_n \sigma} \cdot |\chi(\tau, \nu)|}{\sigma_d^2} \right]$$



$$p(\xi | H_1) = p_{\xi_0}(\xi | H_1) * p_{\xi_1}(\xi | H_1) * \cdots * p_{\xi_{N-1}}(\xi | H_1)$$

$$p_{\xi}(\xi | \sigma; H_1) = \frac{1}{\sigma_d^2} \left(\frac{\xi}{N \sigma \cdot |\chi(\tau, \nu)|^2} \right)^{\frac{N-1}{2}} \exp \left\{ -\frac{\xi + N \sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2} \right\} \cdot I_{N-1} \left[\frac{2\sqrt{\xi N \sigma} \cdot |\chi(\tau, \nu)|}{\sigma_d^2} \right]$$

$$SNR = \frac{\sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2}$$

$$P_d(SNR) = \int_{T^2/\sigma_d^2}^{\infty} \left(\frac{t}{N \cdot SNR} \right)^{\frac{N-1}{2}} \exp \{ -t - N \cdot SNR \} \cdot I_{N-1} \left[2\sqrt{t N \cdot SNR} \right] dt$$



Sistemi Radar

Richiamo DDP NCI (II)

X

$$p_{\xi_n}(\xi_n | \sigma; H_1) = \frac{1}{\sigma_d^2} \exp \left\{ -\frac{\xi_n + \sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2} \right\} \cdot I_0 \left[\frac{2\sqrt{\xi_n \sigma} \cdot |\chi(\tau, \nu)|}{\sigma_d^2} \right]$$

$$\int_0^\infty e^{-(x+a^2)} I_0 \left[2a\sqrt{x} \right] dx = 1$$

$$C_{\xi_n}(\omega | H_1) = \int_0^\infty \frac{1}{\sigma_d^2} \exp \left\{ -\frac{\xi_n + \sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2} \right\} \cdot I_0 \left[\frac{2\sqrt{\xi_n \sigma} \cdot |\chi(\tau, \nu)|}{\sigma_d^2} \right] e^{-j\omega \xi_n} d\xi_n =$$

$$= \frac{1}{\sigma_d^2} \exp \left\{ -\frac{\sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2} \right\} \int_0^\infty \exp \left\{ -\left(\frac{1}{\sigma_d^2} + j\omega \right) \xi_n \right\} \cdot I_0 \left[\frac{2\sqrt{\xi_n \sigma} \cdot |\chi(\tau, \nu)|}{\sigma_d^2} \right] d\xi_n =$$

$$= \frac{1}{\sigma_d^2} \exp \left\{ -\frac{\sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2} \right\} \frac{1}{\frac{1}{\sigma_d^2} + j\omega} \int_0^\infty \exp \{-x\} \cdot I_0 \left[\frac{2\sqrt{\sigma} \cdot |\chi(\tau, \nu)|}{\sigma_d^2 \sqrt{\frac{1}{\sigma_d^2} + j\omega}} \sqrt{x} \right] dx =$$

$$= \frac{1}{\sigma_d^2} \exp \left\{ -\frac{\sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2} \right\} \frac{1}{\frac{1}{\sigma_d^2} + j\omega} \exp \left\{ -\frac{\sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^4 \left(\frac{1}{\sigma_d^2} + j\omega \right)} \right\} =$$

$$= \frac{1}{1 + j\omega \sigma_d^2} \exp \left\{ -\frac{\sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2 (1 + j\omega \sigma_d^2)} \right\}$$

Sistemi Radar

Richiamo DDP NCI (III)

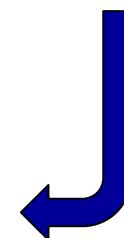
$$p(\xi|H_1) = p_{\xi_0}(\xi|H_1) * p_{\xi_1}(\xi|H_1) * \dots * p_{\xi_{N-1}}(\xi|H_1)$$



$$C_\xi(\omega|H_1) = \prod_{n=0}^{N-1} C_{\xi_n}(\omega|H_1) = C_{\xi_n}^N(\omega|H_1) = \frac{1}{(1 + \omega \sigma_d^2)^N} \exp \left\{ \frac{N \sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2 (1 + j \omega \sigma_d^2)} \right\}$$



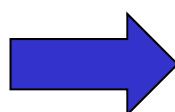
$$p_\xi(\xi|\sigma; H_1) = \frac{1}{\sigma_d^2} \left(\frac{\xi}{N \sigma \cdot |\chi(\tau, \nu)|^2} \right)^{\frac{N-1}{2}} \exp \left\{ -\frac{\xi + N \sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2} \right\} \cdot I_{N-1} \left[\frac{2 \sqrt{\xi N \sigma} \cdot |\chi(\tau, \nu)|}{\sigma_d^2} \right]$$



Richiamo DDP NCI (IV)

$$P_d(\sigma) = \text{Prob}\left\{\xi > T^2 \mid \sigma; H_1\right\} = \int_{T^2}^{\infty} p_{\xi}(\xi \mid \sigma; H_1) d\xi$$

$$\begin{aligned} P_d(\sigma) &= \int_{T^2}^{\infty} \frac{1}{\sigma_d^2} \left(\frac{\xi}{N\sigma \cdot |\chi(\tau, \nu)|^2} \right)^{\frac{N-1}{2}} \exp \left\{ -\frac{\xi + N\sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2} \right\} \cdot I_{N-1} \left[\frac{2\sqrt{\xi N\sigma} \cdot |\chi(\tau, \nu)|}{\sigma_d^2} \right] d\xi = \\ &= \int_{T^2/\sigma_d^2}^{\infty} \left(\frac{t \sigma_d^2}{N\sigma \cdot |\chi(\tau, \nu)|^2} \right)^{\frac{N-1}{2}} \exp \left\{ -t - \frac{N\sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2} \right\} \cdot I_{N-1} \left[\frac{2\sqrt{t N\sigma} \cdot |\chi(\tau, \nu)|}{\sigma_d} \right] dt \end{aligned}$$



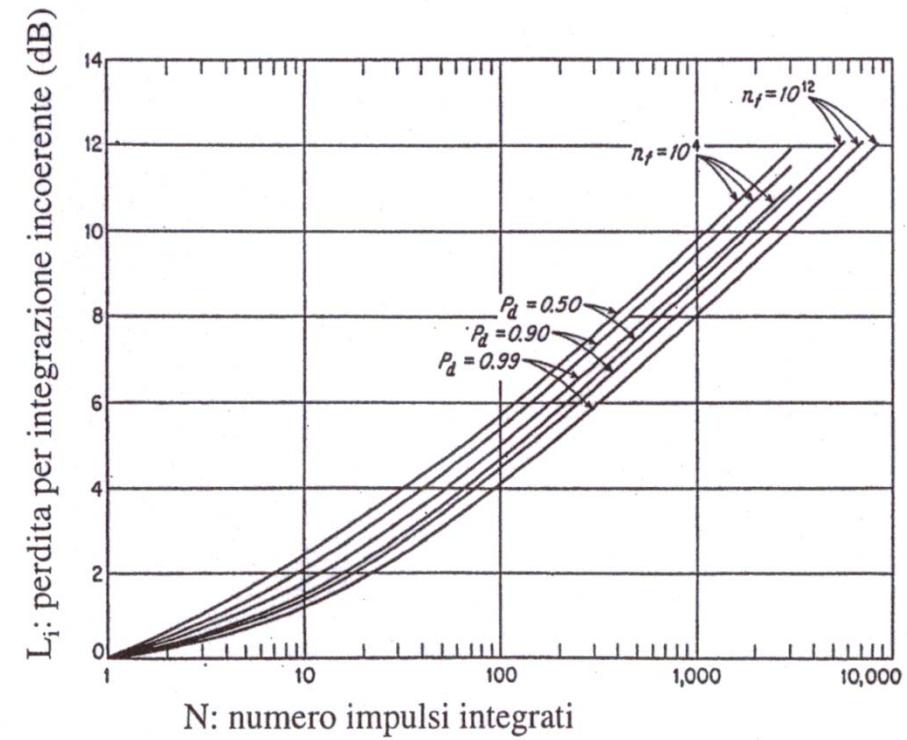
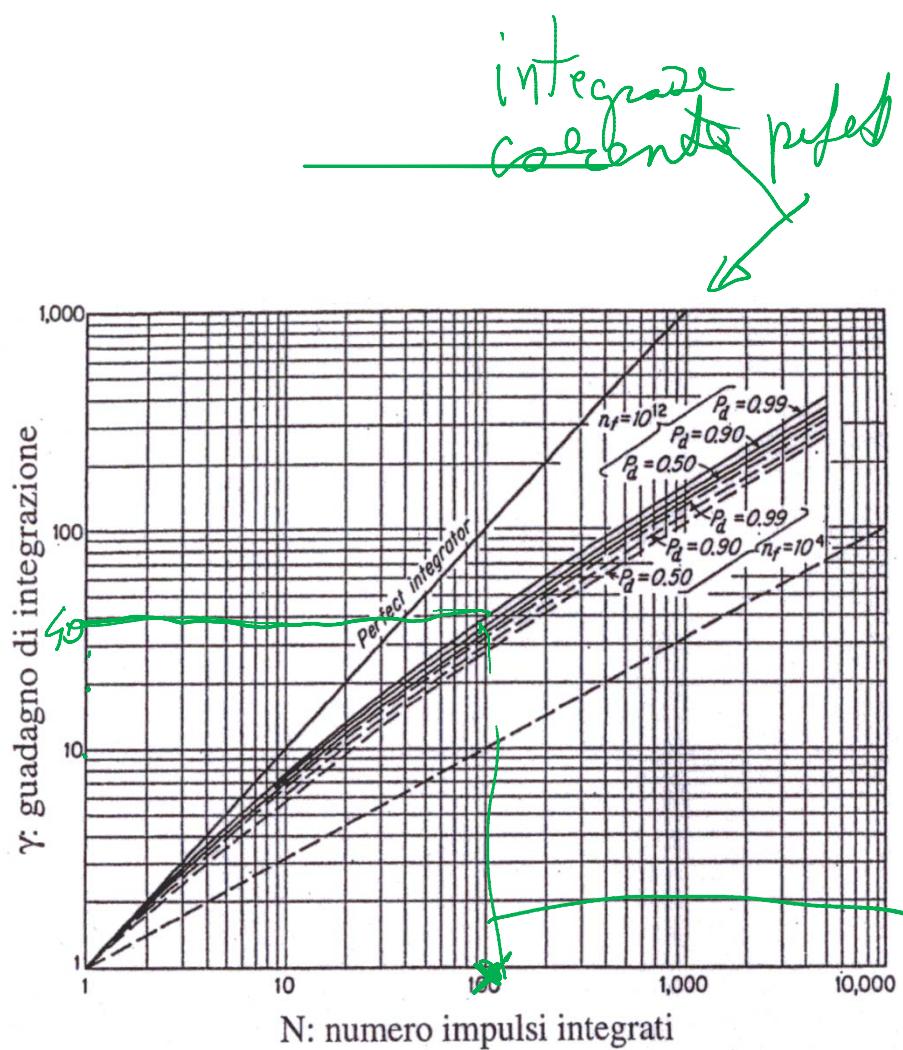
$$P_d(SNR) = \int_{T^2/\sigma_d^2}^{\infty} \left(\frac{t}{N \cdot SNR} \right)^{\frac{N-1}{2}} \exp \left\{ -t - N \cdot SNR \right\} \cdot I_{N-1} \left[2\sqrt{t N \cdot SNR} \right] dt$$



Sistemi Radar

$$SNR = \frac{\sigma \cdot |\chi(\tau, \nu)|^2}{\sigma_d^2}$$

Integrazione incoerente quadratica (II)



$$SNR_{SI} = 13 - 10 \log_{10} 40$$

per $N = 100$

Integrazione incoerente quadratica (III)

Portata radar in presenza di integrazione

$$SNR^N = \frac{1}{\gamma} SNR^I$$

Guadagno di
integrazione

γ

$$L_i = 10 \log_{10} \left(\frac{N}{\gamma} \right)$$

Perdita
integrazione
incoerente rispetto
alla coerente L_i

$$R_{\max} = \left[\frac{E_t G A_e \sigma}{(4\pi)^2 L k T_0 F SNR^*/\gamma} \right]^{1/4}$$

Sistemi Radar

Nel caso di integrazione incoerente quadratica:

- $\gamma \approx N$ per elevati valori di rapporto segnale a rumore;
- $\gamma \approx \sqrt{N}$ per bassi valori di rapporto segnale a disturbo
 - $L_i \approx 0 \text{ dB}$ per elevati valori di rapporto segnale a rumore;
 - $L_i \approx 10 \log_{10}(\sqrt{N})$ per bassi valori di rapporto segnale a disturbo



Il guadagno di
integrazione γ
consente di

Aumentare la portata di $\gamma^{1/4}$ se si lascia
inalterata l'energia trasmessa fissata per il
caso di decisione su singolo impulso

Trasmettere un'energia γ volte inferiore
rispetto a quella fissata per il caso di decisione
su singolo impulso se si lascia inalterata la
portata

Albersheim's Equation

Albersheim's Equation

Albersheim's equation is a closed-form approximation to the SNR required to achieve the specified detection and false alarm probabilities for a nonfluctuating target in independent and identically distributed Gaussian noise. The approximation is valid for a linear detector and is extensible to the noncoherent integration of N samples.

Let

$$A = \ln \frac{0.62}{P_{FA}}$$

and

$$B = \ln \frac{P_D}{1-P_D}$$

where P_{FA} and P_D are the false alarm and detection probabilities.

Albersheim's equation for the required SNR in dB is:

$$\text{SNR} = -5\log_{10} N + [6.2 + 4.54 / \sqrt{N + 0.44}] \log_{10}(A + 0.12AB + 1.7B)$$

where N is the number of noncoherently integrated samples.

where N is the number of noncoherently integrated samples

Sistemi Radar

$\rightarrow N=1 \rightarrow$ lavoro con 1 impulso

$\rightarrow N \rightarrow N \rightarrow$ lavoro con N impulsi in modo incerto

– per integrazione coerente con N impulsi:

metto $N=1$ e sostengo $\log_{10} N$